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USSR Report

ELECTRONICS AND ELECTRICAL ENGINEERING

(FOUO 6/81)



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USSR REPORT
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CONTENTS

ANTENNAS

Suppression of Radiation of a Parabolic Horn Antenna on the E ₀₁ Wave.....	1
Antennas of 'Moskva' Stations.....	7

COMMUNICATIONS, COMMUNICATION EQUIPMENT, RECEIVERS AND TRANSMITTERS,
NETWORKS, RADIO PHYSICS, DATA TRANSMISSION AND PROCESSING, INFORMATION
THEORY

Book Discusses Construction of Long-Distance Communications Cable Lines.....	12
---	----

INSTRUMENTS, MEASURING DEVICES AND TESTERS, METHODS OF MEASURING,
GENERAL EXPERIMENTAL TECHNIQUES

Modulation Superheterodyne Radiometer of the 2-mm Wave Band With a Josephson Frequency Converter at the Input.....	21
---	----

MICROELECTRONICS

Instruments of Functional Semiconductor Electronics.....	29
--	----

PUBLICATIONS, INCLUDING COLLECTIONS OF ABSTRACTS

Book Discusses Design and Optimization Criteria for Semiconductor Devices.....	30
A Concise Dictionary of Radio-Electronics.....	32
Control of Terrestrial Antennas and Radiotelescopes.....	33
Component Structure in Large-Scale Integrated Circuits.....	36

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Current Stabilization Methods and Equipment.....	39
Installation and Set-Up of Industrial Electrical Communication and Alarm Systems.....	46
New Collection Examines Electronics Materials and Equipment.....	51
Optimization and Control in Electronic Vacuum Production.....	53
Photoelectronic Control Systems.....	55
Semi-Markov Processes With a Discrete Set of States.....	58

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ANTENNAS

UDC 621.317.088:621.317.342

SUPPRESSION OF RADIATION OF A PARABOLIC HORN ANTENNA ON THE E_{01} WAVE

Moscow ELEKTROSVYAZ' in Russian No 1, Jan 81 pp 58-61

[Article by A. A. Metrikin and V. G. Petrova, manuscript received 26 March 1980]

[Text] Introduction. Parabolic horn antennas (RPA) are used widely on radio relay lines (RRL) in the ranges of 4 and 6 GHz [1,2]. Circular waveguides with the diameter $2a = 70$ mm are used as feeders. The fundamental wave is H_{11} . In the range of 4 GHz ($2.85 \leq ka \leq 2.49$, $k = 2\pi/\lambda$), along with the fundamental wave H_{11} , an E_{01} having a different group velocity of propagation propagates in the waveguide. In the range of 6 GHz ($4.11 \leq ka \leq 4.55$), E_{01} , E_{11} , H_{21} , H_{01} waves also propagate in a 70 mm--diameter waveguide.

The greatest influence on the operation of RRL within the range of 6 GHz is exerted by E_{01} and E_{11} waves which have a longitudinal component of the electric field ($E_z \neq 0$). The mechanism of the formation of parallel flows due to parasitic waves is examined in [2]. Let us only mention that during the formation of parallel flows the main RRL parameter -- nonuniformity of the group delay time changes and, as a result of this, there occurs a nonlinear transition noise in the telephone channels.

The excitation of parasitic waves E_{01} , E_{11} and others is caused both by nonuniformities of the waveguide channel (steps, ellipticity, bends of the channel), and by inaccurate orientation of the RPA during the reception or transmission of energy, i.e., when the RPA is turned away from the main direction of energy reception.

Below are given the results of experimental studies on the operation effectiveness of an absorber of the parasitic wave E_{01} installed in a circular waveguide. The effectiveness of suppression is evaluated by the decrease in the energy level of the E_{01} wave in the RPA.

Experimental Conditions. Studies were conducted on an RPA-2P-2 antenna (Figure 1) in the reception mode. The antenna was installed on a tower 20 meters high. The transmitting antenna was excited on an H_{11} wave with horizontal or vertical polarization of the field. For receiving the E_{01} wave, a E_{01} -wave exciter (Figure 2) was connected to the waveguide. The exciter is a coaxial transition with an oscillator oriented along the axis of the waveguide where the field E_z of the E_{01} wave is maximal. The connection of the exciter with the RPA is shown in Figure 1b.

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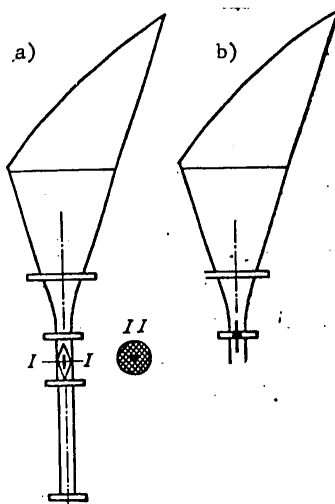


Figure 1

A glass rod with an oxide film was used as an absorber of the E_{01} wave over the range of 4 GHz. The length of the rod was 200 mm, its diameter was 2 mm, and the direct-current resistance was 2000 ohms. The rod was installed along the axis of the waveguide with a diameter of 70 mm in a foam plastic holder (Figure 3a). The design of the absorber was somewhat different for the range of 6 GHz. We used four glass rods (resistors) which were installed at the points where E_z was maximal for the E_{11} wave. One pair was intended for suppressing the E_{11} wave with horizontal polarization of the field, the other pair -- with vertical polarization. The two pairs of glass rods absorb, although less effectively, also the E_{01} wave. The glass rods in this case were fastened with capron threads (Figure 3b).

Results obtained. Directional patterns of the RPA were obtained in two frequency ranges ($f = 3.68$ and 5.86 GHz) in the horizontal plane on the E_{01} wave for the cases with and without the absorber. The energy level of the E_{01} wave was read from the energy level of the H_{11} wave received from the principal direction.

The directional patterns of the RPA on the E_{01} wave at a frequency $f = 3.68$ GHz for horizontal and vertical polarizations of the field during transmission are shown in Figures 4 and 5, respectively (solid line -- without the absorber, broken line -- with the absorber). As can be seen from Figure 4, the attenuation of the field in the main direction is ~ 30 dB in relation to the H_{11} wave. The maximum level of the field on the E_{01} wave is minus 8 dB at the angles of ± 1.25 degrees. When the absorber of the E_{01} wave is installed in the wave guide, the field level of this wave attenuates considerably (in this case by 36 dB).

In the case of the vertical polarization of the field (Figure 5), due to the absence of symmetry along the electric field, there is no dip on the E_{01} wave during the reception from the main direction. The level of the field excited in the receiving

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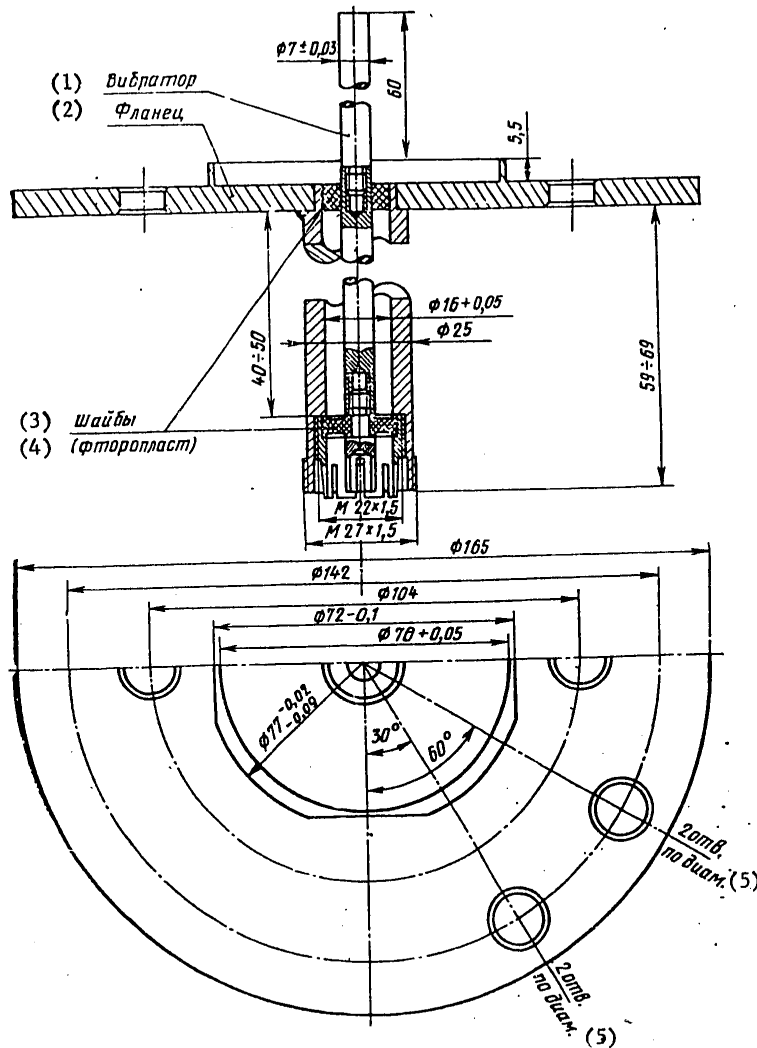


Figure 2

Key: 1. Oscillator
2. Flange

3. Washers
4. (Fluoroplastic)

5. 2 Apertures along the Diameter

antenna on the E_{01} wave is considerably lower than the level of the field in the case of horizontal polarization. In this case, the absorber noticeably attenuates the level of the side lobes (of the order of 20 dB) of the E_{01} wave.

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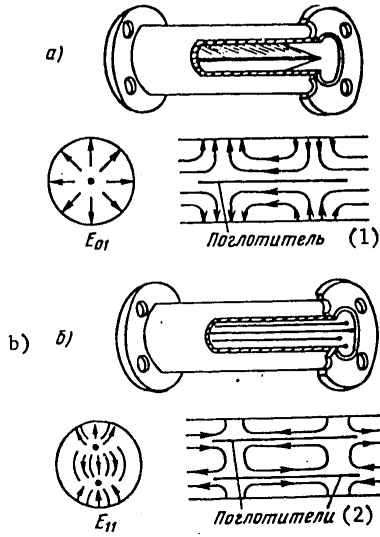


Figure 3

Key: 1. Absorber
2. Absorbers

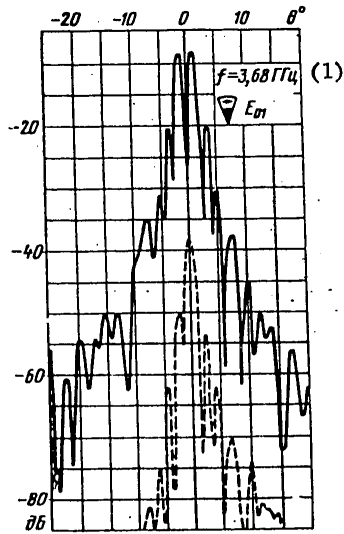


Figure 4
Key: 1. GHz

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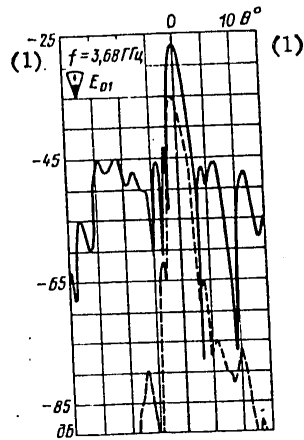


Figure 5.
Key: 1. GHz

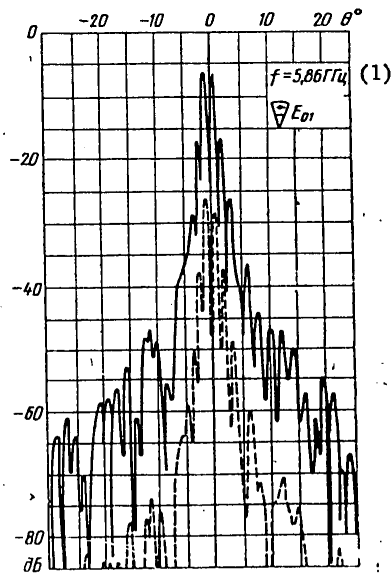


Figure 6
Key: 1. GHz

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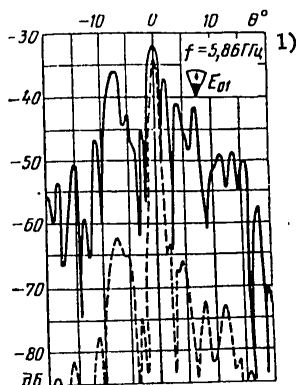


Figure 7

Key: 1. GHz

Analogous directional patterns for the frequency $f = 5.86$ GHz are shown in Figures 6 and 7. It can be seen from them that the installation of the absorber of the E_{01} wave in the circular waveguide considerably lowers its level.

A good symmetry of the directional patterns of the antenna on the E_{01} wave with a deep trough in the principal direction during horizontal polarization of the incident field can be used for accurate orientation of the antenna in the horizontal plane at a minimal level of the parasitic wave E_{01} . Experience shows that with this method the RPA can be oriented accurate to $\pm 5'$. In the case of the orientation in the vertical plane, the incident field must have vertical polarization. Let us note that the method of the orientation of antennas on the principal wave H_{11} which was used until recently does not make it possible to align the antenna in the principal direction of radiation with such a high degree of accuracy.

Conclusions. 1. Absorbers lower the level of the E_{01} wave to minus 25 dB (maximum value) instead of 7-8 dB (maximum value).

2. For accurate alignment of an RPA, it is recommended to use the method of orientation by the E_{01} wave at its minimal level.

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ANTENNAS OF 'MOSKVA' STATIONS

Moscow ELEKTROSVYAZ ' in Russian No 1, Jan 81 pp 61-64

[Article by A. G. Kvitko and A. M. Pokras: "Antennas of 'Moskva' Stations", manuscript received 13 May 1980]

[Text] The role and place of the satellite television broadcast system "Moskva" were defined in [1,2]. The same works gave the basic characteristics of the equipment of the receiving ground station (ZS). The purpose of this work is a more detailed description of the antennas of the stationary and transportable stations (Figure 1).

The ZS antenna had the following requirements: gain coefficient (KU) in the sector of angles ± 1 degree not lower than 35 dB, noise temperature during the direction of the beam at an angular altitude of 5 degrees not over 70 K, level of the first side lobe not over minus 20 dB, work at circular polarization with an ellipticity factor (KE) of not less than 0.7, manual coarse adjustment of the antenna beam within the limits of ± 90 degrees by the azimuth and 0-60 degrees by the angular altitude, smooth manual adjustment by the azimuth and angular elevation ± 5 degrees, the possibility of adding a servo drive for operation in the mode of automatic tracking with smooth beam guidance by two axes within the limits of ± 2.5 degrees, simplicity and technological effectiveness of the design.

The antenna used a series produced mirror 2.5 meters in diameter with an angular aperture of 160 degrees and a focal distance of 750 mm. With the above-mentioned diameter, it is possible to fulfill the requirement for a low level of side lobes, ensuring the prescribed gain. The lowering of the level of side radiation is achieved due to a rapid drop of the amplitude of the field toward the edges of the mirror aperture at a low blockage of the aperture by the feed. Therefore, the irradiation of the mirror is designed as a single-mirror circuit. The small dimensions of the mirror eliminated the possibility of using a two-mirror irradiation circuit, because in the range of 4 GHz the minimum size of the counter-reflector would exceed 0.5 m and the aperture blockage would become impermissibly great.

The requirement for operation with circular polarization made it practical to use a simple spiral feed (Figure 2). It is a double-screw logarithmic spiral excited by a coaxial balun. The latter is designed in the form of two half-wave diametrically opposite longitudinal slits in the external conductor of the coaxial line. One of the sides of the slits in the middle is connected by a short-circuiting jumper with

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the central conductor. A circular short-circuiting jumper is installed at the end of the balun.

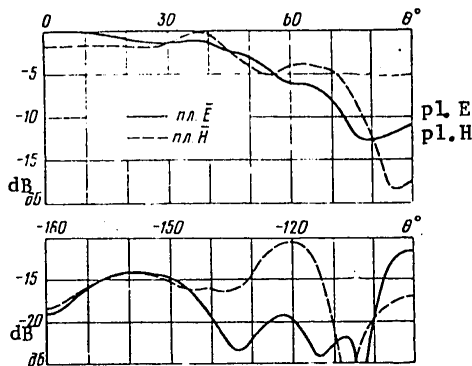


Figure 3

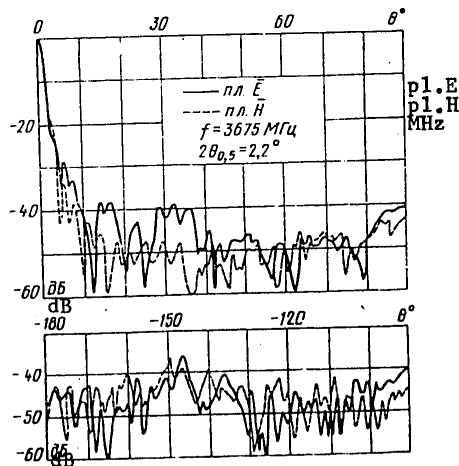


Figure 4

In order to match the effective resistance of the balun loaded with spirals and the coaxial line with a wave resistance of 50 ohms, a transformer in the form of a quarter-wave insert was introduced in the central conductor. The reactive components are matched by means of adjustment screws passing through the external coaxial conductor which further passes through the side waveguide-to-coaxial adapter to the waveguide with a cross section of 58X25 mm. The dimensions of the spirals are selected in such a way that in the direction of the edges of the mirror (angles ± 80 degrees) the irradiation level along the directional pattern (DN) of the feed

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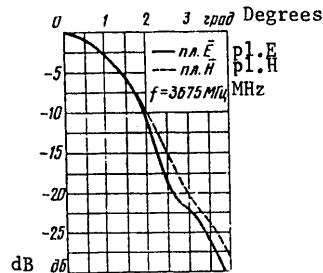


Figure 5

fell by 11.5 dB. With the consideration of the difference of the routes from the phase center of the feed to the center and the edges of the mirror, the drop in the excitation to the edges of the aperture reached minus 16 dB. In this case, it was possible to obtain the ellipticity of the coefficient of the antenna higher than was prescribed.

The directional patterns of the feed and the antenna are given in Figures 3, 4, and 5 (due to the symmetry, halves of the DN are given). The level of the first side lobe does not exceed minus 20 dB. Measurements showed that the antenna had the following characteristics: KU at a frequency of 3675 MHz was 37.5 dB (KIP 0.61) and in the sector ± 1 degree not less than the prescribed 35 dB, KE did not exceed 0.85, KSV [standing wave ratio] at the input of the feed in the working range was not over 1.2.

In order to reduce the power of the electric servo drive, the orientation of the beam in the automatic-tracking variant of the antenna (Figure 6) is done without turning the mirror. As it follows from the DN shown in Figure 4, the prescribed limits of guidance to ± 2.5 degrees are approximately one width of the DN with respect to the level of half-power. Such a turn of the beam can be obtained by transverse shifting of the feed by 40 mm with a gain loss of less than 0.5 dB [3]. In order to simplify the design, the carrier pipe of the feed was secured on two hinges located in the vicinity of the top of the mirror which made it possible to move the feed in two orthogonal planes by turning the carrier pipe. Since the distance from the hinges to the phase center of the feed was about 0.8 m, the circular arc practically does not differ from the chord when the swing angle is small.

The swinging device of the feed is equipped with an electric servo drive. The beam is moved with a discreteness of 12' on commands from a two-coordinate extremal automatic device which performs consecutive searching of the extremum along two axes. Structurally, discreteness is achieved with the aid of the combination of reducers and a "Maltese cross" which eliminates uncontrolled turns due to the "rundown" of the motor after de-energization.

The stand on which the antenna for ZS is installed is stationary (Figure 7) and is a combination of welded frames connected with hinges and joined with rods whose lengths can be regulated with the aid of turnbuckles 5, 6. The location points of the hinges are selected on the basis of the prescribed angular coordinates of the

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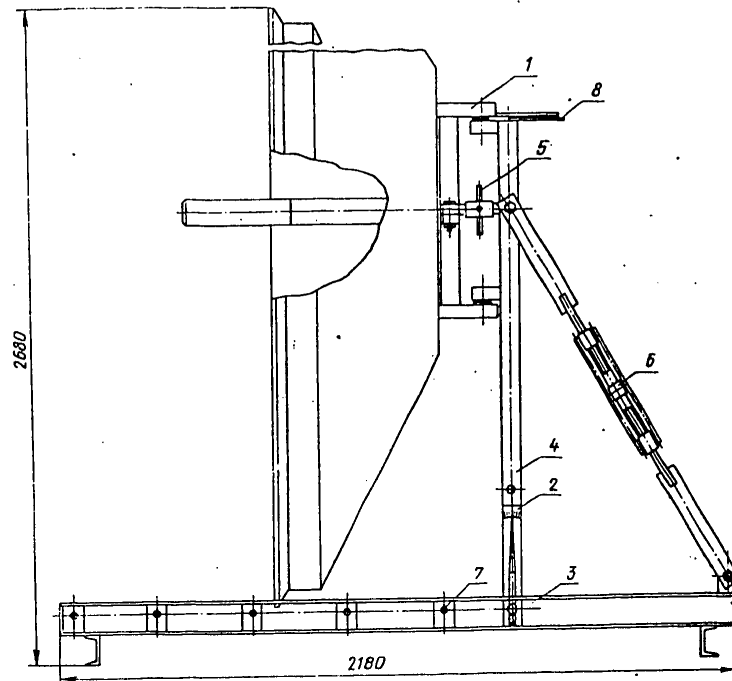


Figure 7

ISZ [artificial earth satellite]. For this purpose, holes 7 are drilled in the walls of the frames. The bolts inserted into them connect the horizontal frame 3 with frame 4. The horizontal frame is positioned approximately along the prescribed azimuth. On frame 4, also on hinges, is attached frame 1 to which the mirror is bolted. The mirror turns smoothly with the aid of turnbuckles along the azimuth at ± 7.5 degrees and turns up and down at ± 5 degrees. The readings of the angles are taken from scales 2 and 8. In this manner, the beam of the antenna is guided to the ISZ. If an electric drive is installed on the antenna, further ISZ tracking, if necessary, can be achieved by the swinging of the feed.

The stand for the antenna of a transportable station is shown in Figure 8. Its substantial differences from the stand for a stationary variant are explained by the following circumstances: the antenna arrangement must be compact and allow for the possibility of setting the mirror in the transportation position (zenith). Accordingly, the limits of the angular elevation turns of the mirror are widened from 0 to 90 degrees, which is accomplished with the aid of sectional pull rods 1. Smooth manual angular elevation adjustment is done within the limits of ± 10 degrees with the device 2 (a motion screw and a nut) fixed in the required part of the sectional pull rods. The chassis of the trailer of the station is preset in the position along the prescribed azimuth. Smooth adjustment is done manually with turnbuckle 3. Since the antenna is rigidly fixed to the chassis on which the cabin is situated,

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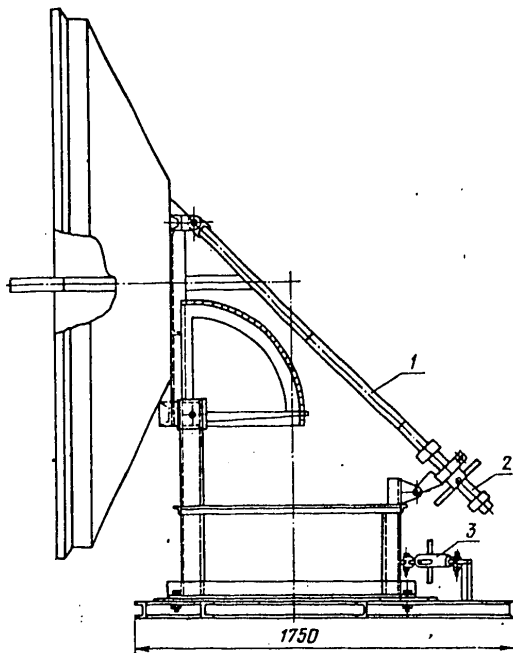


Figure 8

the limits of the movement of the mirror along the azimuth were increased to ± 10 degrees.

The designs of the stands of both types allow for the possibility of installing an electric drive for swinging the feed.

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BOOK DISCUSSES CONSTRUCTION OF LONG-DISTANCE COMMUNICATIONS CABLE LINES

Moscow UKAZANIYA PO STROITEL'STVU MEZH DUGORODNYKH KABEL'NYKH LINIY
SVYAZI in Russian 1972 pp 3, 568-575

[Foreword and table of contents from book "Instructions for the Construction of
Long-Distance Communications Cable Lines", Izdatel'stvo "Svyaz'"]

FOREWORD

/Text/ Existing statutes on rules, directions, instructions, technical conditions, state standards and so forth were used as the basis for drawing up these Instructions for the construction of long-distance communications cable lines.

The content of these Instructions corresponds to the present-day level of equipment, technology and organization of the construction of long-distance communications cable lines. The Instructions incorporate experience that has been accumulated in designing, constructing and operating long-distance communications cable lines, the industrialization and mechanization of construction and the use of leading labor methods.

These instructions must be observed by construction, design, operational and other organizations of the USSR Ministry of Communications. The publication of these Instructions renders the "Rules for the construction of long-distance communications cable lines", published in 1958, null and void.

These Instructions were developed by the Giprosvyaz' /State Institute for the Surveying and Designing of Communications Installations/ Institute, reviewed by an expert commission, revised and edited by the Orgtekhsvyaz'stroy /communications construction/ Trust. The development, delivering of an expert opinion and revision work involved leading specialists from construction, designing, operational, and scientific-research organizations of the USSR Ministry of Communications.

Chapters 1, 2, and 3 were compiled by D.A. Baron, Chapter 4 by N.S. Oreshkin and A.D. Isayevich, Chapter 5 by A.D. Isayevich, Chapter 6 by Ye.A. Kozyreva and N.A. Osipov, Chapter 7 by D.A. Baron and B.I. Gershman, Chapter 8 by A.Ye. Slavina, Chapter 9 by N.S. Oreshkin and A.D. Isayevich, Chapter 10 by K.P. Fomin, A.D. Isayevich and N.K.

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Gorodetskaya, chapters 11, 12 and 13 by V.N. Akulenk. Chapter 14 by Yu.G. Fedorov, chapters 15 and 16 by V.A. Shipilov and D.M. Samorukov, Chapter 17 by F.L. Letichevskiy, D.Yu. Brodskaya, A.Ye. Slavina and D.I. Kifer, and the appendices were compiled by M.P. Solov'ev and D.A. Baron.

The Instructions were reviewed by an expert commission of the USSR Ministry of Communications. The commission members were: N.S. Glagolevskiy (chairman), M.K. Titov (deputy chairman), G.S. Mizheritskiy, M.T. Nefedov, A.N. Golubyatnikova, V.P. Skripnik, V.N. Alekseyev, A.Ye. Yakovich, P.P. Dorozhko, M.D. Mel'nikova, V.I. Chesnokova. The expert conclusion with specific recommendations was prepared by a working group of the commission, comprised of: D.A. Baron, A.D. Isayevich, G.S. Mizheritskiy and V.P. Skripnik.

All comments and suggestions concerning the Instructions should be sent to the SVYAZ' Publishing House (Moscow-Center, Chistoprudnyy Boulevard, 2).

Main Administration of Capital Construction of the USSR Ministry of Communications, Giprosvyaz' and Orgtekhsvyaz'stroy.

TABLE OF CONTENTS	PAGE
Foreword	3
Chapter 1. General Information on Long-Distance Communications Cable Lines	4
Classification, composition and basic requirements for long-distance communications cable lines	4
Types and designs of long-distance cables	5
Cable markings	9
Purpose of cables	14
Basic theories on designing and the requirements for design documentation	16
Chapter 2. Basic Elements of the Technology and Organization of the Construction of Long-Distance Communications Cable Lines	19
General information	19
Organization of performing of preparation work	20
The draft for organizing the construction	22
The PPR (draft for performing work)	26
Preparation of equipment	34
Preparation of motor transport	35
Material-technical support	37
Preparation of measuring equipment	38
Providing construction with non-standard structures	39
Determining need for work force	40
Organizing production subelements	41

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Responsibilities and composition of production subelements 42
 Basic responsibilities of participants 48
 Responsibilities of construction customers and contract organizations 50

Chapter 3. Work at the Cable Sites 53

Siting the cable locations on the route 53
 Acceptance of cable drums, equipment and fittings at unloading points 54
 Equipment of motor vehicles and transporting cable 56
 Testing cable and equipment at cable sites 58
 Repairing cable and equipment 60
 Grouping cables and coil boxes in front of laying work 61

Chapter 4. Laying Underground Cables 65

Route layout 65
 Cable laying with a cable-laying machine 66
 Digging trenches and ditches 67
 Laying pipelines at highway and railroad crossings 72
 Unwinding cable and laying it in prepared trench 73
 Back filling of trenches 75
 Anti-erosion measures on route of mainline cables 75
 Handing the laid cable over for installation 78
 Fixing the route of the laid cable 79
 Special features of laying cables in winter conditions 81

Chapter 5. Laying Cables in Permafrost Regions 82

General instructions 82
 Conditions of laying cables in permafrost regions 83
 Methods of laying cables in permafrost regions 84
 Recommendations for organizing and mechanizing work 88

Chapter 6. Laying Cables Accross Water Barriers 90

General instructions 90
 Preparation work 94
 Ways to lay and deepen cable 96
 Laying cable with a cutter cable-laying machine 96
 Laying cable from a floating platform or manually into prepared
 trenches and developing the trenches 99
 Back filling underwater trenches 107
 Securing underwater cables at the shorelines 107
 Protection of underwater cable crossings 108

Chapter 7. Installation Work 109

Estimating procedure on cable mainlines 109
 Assembling the installation-measuring column 112
 Preparation measures 114
 Cable installation 115

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Receiving laid cable for installation	115
Composition of work for installing sleeves	117
Preparing the work place	118
Checking the cables before installation	120
Preparing cable ends prior to installing the sleeves	122
Installing direct sleeves	128
Installing fork sleeves	141
Installing symmetrical sleeves	145
Installing condenser sleeves	149
Placement and installation of coil boxes	150
Placement and installation of extension pieces	154
Installation of joint sleeves of high frequency cables of a symmetrical design	155
Protecting installed sleeve from mechanical damage, back filling of trenches, and fixing of the location of the sleeve and cables .	158
Special features of installing screened pairs and units of four	161
Installation of cables with coaxial pairs	163
General information	163
Installation of type KBM (K,G)-4 cables with coaxial pairs 2,6/9,4	166
Preparing cable ends and coaxial pairs for installation	166
Splicing coaxial pairs	168
Special features of splicing symmetrical units of four	171
Installing coils and signalizers for reducing pressure in cable sleeves, type KM-4	172
Packing the splice and soldering the lead sleeve	172
Installation of the type 5/18 coaxial pair with disk insulation....	173
Installation of the type MKTS-4 small coaxial cable	175
Special tools, parts and materials used in installing type MKTS-4 cable	175
Preparing cable ends for installation	179
Splicing coaxial pairs	181
Splicing duty pairs and the control core	182
Installation of cables in an aluminum casing	183
General instructions	183
Preparing cable ends for installation	183
Installing sleeves using hot solder	185
Installing sleeves using VK-9 adhesive	186
Restoring insulation covers	187
Installing air-tight sleeves	192
Layout of lead-outs for the KIP on construction length of cable	192
Soldering the cable ends	192

FOR OFFICIAL USE ONLY

Installation of cables in plastic casing	193
General instructions	193
Materials used in installation using the cold method on type MKPV cable	194
Installation of direct sleeves	196
Installation of symmetrical sleeves	200
Installation of clad cable, type MKPVB 1X4X1.2	203
Installation of MKPV cable 1X4X1.2 using heating linings and ribbon glass	203
Preparing cable for installation	203
Installation of direct sleeves	205
Installation of symmetrical sleeves	207
Installation of MKTP-4 cable using heating linings and rib- bon glass	207
Installation of switching sleeves on cable ends in a type MKTP-4 plastic casing with cables in a lead casing, type MKTSB-4	213
General instructions	213
Splicing the end of a MKTP-4 (Drawing 7.126) cable	213
Splicing the end of a MKTSB-4 (Drawing 7.81) cable	214
Installation of sleeves	215
Installation of type MKPTSH 1X4X1.2 cable	216
Installation of special design cables	216
General information	216
Installation of MKSEBv cable with a plastic hose	216
Installation of type MKSEBv cable with plastic ribbons	220
Installation of type MKBKMB cable	221
Installation of type MKPLMB cable	223
Installation of cables on river crossings	224
General information	224
Sleeveless connection of wire casing on underwater cables ..	225
Installation of coupling protective sleeve with pear-shaped linings	231
Installation of marine cables	233
General instructions	233
Installation of underwater coaxial cable type KPK-5/18	234
Splicing of cable ends	234
Splicing of internal wire	234
Restoration of polyethylene insulation	235
Splicing of exterior wire	238
Restoration of protective covers	239
Splicing the casing	240

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Installation of type KPEB-5/18 cable	241
Splicing an underwater KPK-5/18 cable to an underground KPEB-5/18 cable	243
Installation of underwater symmetrical type SEPK-4 cable ...	244
Chapter 8. Checking the Hermetic Seal of Cable Casings and Putting Them Under Gas Pressure	247
Systems for maintaining cables under gas pressure	247
Equipment used in systems for maintaining cables under pressure .	250
Gas-tight sleeves	250
AKOU (automatic control-drying device)	251
Compressor-signal device (KSU)	253
Air transfer shield (SHCHPV)	255
Control-drying device (KOU-1 and KOU-2)	255
Control shield	257
Drying device	257
Humidity indicator	259
Compressed air tanks	259
Reducers	260
Compressors	260
Regeneration device (RU)	262
Complete set of equipment needed to determine precisely the place where the hermetic seal of a cable casing has been broken (KO)	263
Mobile compressor station (PKS)	264
Field charging carbonic acid station (PZUS)	264
Installation of AKOU	264
Installation of gas-tight symmetrical sleeves (GMS and MG)..	266
Installation of gas-tight terminal coaxial sleeve (OGKM) ...	270
General instructions on positioning cables under gas pressure and checking their hermetical seal	274
Testing the hermetical seal of cable casing in construction lengths	277
Testing the hermetical seal of connecting sleeves, coil boxes and extenders	279
Testing the hermetical seal of cable casings in steps, sections and on installed amplifier sectors	279
Testing the hermetical seal of input terminal devices	281
Filling out records on hermetical seal testing	283
Basic theories on the maintenance of equipment for keeping cable under pressure, which is used in construction	289
Finding breaks in the hermetic seal of cable casings	290
Chapter 9. Equipment for Cable Inputs and the Installation of Terminal Devices	295
General information	295
Unsoldering long distance cables on distributing cables and the installation of branched sleeves	299

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Installation of boxes. Layout of gas-tight and electrically insulated sleeves	301
Special features of cable input with aluminum casing	306
Input of symmetrical cables into underground NUP /unattended amplification point/	309
Input of symmetrical cables into above-ground NUP	313
Layout of input of coaxial cables into underground NUP	319
Layout of input of cables MKP, MKV, KSPP, VTSP into NUP k-24p ...	325
Input of symmetrical cables multiplexed by the K-60p equipment, into an underground NUP	332
 Chapter 10. Electrical Measurements and Cable Symmetry	 335
General theories	335
Symmetry and electrical measurements of high-frequency symmetrical cables	341
Electrical measurements of coaxial mainline cables type KMP (G,K)-4	358
Electrical measurements of small coaxial cables, MKTP (B)-4, MKTS (B, BP)-4, MKTAP (B, BP)-4	368
Symmetry and electrical measurements of low-frequency cables	373
Symmetry and electrical measurements of cables on branch lines of mainlines	387
Symmetry and electrical measurements of cable insertions into aerial communications lines	387
Instruments for measuring long-distance cable lines	387
 Chapter 11. Protecting Communications Cables from Corrosion	 394
General instructions	394
Equipment for control-measuring points	400
Measuring potentials on cable casing	406
Non-polarized copper sulfate electrodes	410
Equipment for drainage protection of cables	411
Equipment for cathode protection of cables	415
Layout of insulating and electrical insulating sleeves	418
Equipment for the cathodic protection of cables	419
Layout and installation of packed cathodes, type PM10 U/2	422
Protecting metal NUP tanks from corrosion	423
Resoldering cable casings	424
 Chapter 12. Protecting Communications Lines from Lightning	 426
General instructions	426
Protective measures	428
Laying and splicing protective wires	430
Protecting cables using aerial communications lines	434
 Chapter 13. Groundings	 439
Grounding thunder storm protection devices	439
Grounding the NUP	442

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Grounding in soils with a high relative resistance	447
Grounding in permafrost soils	449
Chapter 14. The Construction of Unattended Amplifier Points (NUP)	451
General information	451
Above-ground NUPs	451
NUP design	451
NUP for symmetrical cable 4X4 or 7X4 multiplexed by the K-60 equipment	452
NUP for the KMB-4 coaxial cable multiplexed by the K-1920 equipment	453
Underground NUPs	453
NUP design	453
NUP for symmetrical 4X4 or 7X4 multiplexed by the K-60 equipment	453
NUP for coaxial cable, KMB-4, multiplexed by the K-1920 equipment	458
NUP for symmetrical cable 4X4 multiplexed by the K-60P (SHIP) equipment	460
NUP for small coaxial cable multiplexed by K-300 equipment ..	461
NUP for symmetrical cable 1X4X1.2 multiplexed by K-60p-3 equipment	462
NUP for symmetrical cable 1X4X1.2 multiplexed by type K-24p-2 equipment	465
Chapter 15. Execution Documentation for Long-Distance Cable Line Construction Sites	467
General instructions	467
Composition of the executive documentation	467
Procedure for filling out and formulating executive documentation	480
Chapter 16. Turning Long-Distance Communications Cable Lines Over for Operation	482
Interim certification of work by representatives of the customer .	482
Certificates for concealed work.....	483
Turning individual construction projects over for operation	484
Chapter 17. Equipment Used in the Construction of Long-Distance Communications Cable Lines	494
General instructions	494
Equipment for rigging cuts	494
Equipment for route layout and back filling trenches	498
Equipment for loading and unloading work	498
Motor vehicles, tractors and prime mover equipment	498
Cable laying equipment of the cutter type	501

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Cutters for cutter-type cable-laying machines	513
Excavators	523
Machine for arranging the contours for grounding the MSKZ-2	527
Equipment for non-trench laying of cable pipe lines	529
Small mechanization equipment and auxiliary equipment	531
Housing and production vans	536
Appendix 1. Symmetrical low-frequency cables (GOST-5008-60)	541
Appendix 2. Low-frequency combined cables (GOST-5008-60)	544
Appendix 3. Symmetrical high-frequency cables	547
Appendix 4. Coaxial cables	549
Appendix 5. Sample list of tools, attachments and equipment used by welding brigades in cable installation work	550
Appendix 6. Sample list of materials and fittings used in the construction of long-distance communications cable lines	553
Appendix 7. Characteristics of individual materials	557
Appendix 8. Drums for cable and wire (Drawing P.7.4)	564
Appendix 9. Correlations between nepers and decibels	566

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INSTRUMENTS, MEASURING DEVICES AND TESTERS, METHODS
OF MEASURING, GENERAL EXPERIMENTAL TECHNIQUES

UDC 621.396.621.53.029.65:537.312.62

MODULATION SUPERHETERODYNE RADIOMETER OF THE 2-mm WAVE BAND WITH A JOSEPHSON FREQUENCY CONVERTER AT THE INPUT

Moscow RADIOTEKHNIKA I ELEKTRONIKA in Russian Vol 26, No 1, Jan 81 (manuscript received 4 Dec 79) pp 167-172

[Article by V. S. Ablyazov, S. A. Andreyev, A. N. Vystavkin, V. N. Gubankov, V. P. D'yakov, A. P. Zhukov, A. G. Kislyakov, K. A. Rulev, M. A. Tarasov, S. Yu. Turygin, V. I. Chernyshev]

[Text] This is a report on successful tests of a Josephson superheterodyne radiometer of the 2-mm wave band on the RT-25 X 2 radio telescope of the IPF [Institute of Applied Physics] of the USSR Academy of Sciences. The results obtained make it possible to refine the determination of the brightness temperature of the sun in this band.

Introduction

Although superheterodyne radiometers of the shortwave part of the mm-wavelength band are inferior to crystal video radiometers with respect to the level of the threshold signal ΔT_{f1} [1], they cannot be replaced by the latter in studying spectral characteristics of radiation. At the present time they are actually the only receivers of radiation in this band which preserve information on the phase of the signal. In recent years, the problem of increasing the sensitivity of superheterodyne radiometers has aroused great interest in Josephson's down converters of frequency [2,3,4]. These works presented a detailed study of the operation mechanism of such mixers and demonstrated their high potential sensitivity. However, the laboratory models of mixers used there could not be used in real radiometers due to imperfections in the design.

This work reports on the design, parameters, and results of field tests of a superheterodyne radiometer of the 2-mm wavelength band with a Josephson frequency converter at the input developed by the IRE [Institute of Radio Engineering and Electronics] of the USSR Academy of Sciences. Field tests of the radiometer were conducted on the RT-25 X 2 radiotelescope of the IPF of the USSR Academy of Sciences (Gor'kiy).

1. Design of the Radiometer

The radiometer has a modulation circuit with a pattern switch at the input [1] (direct losses 0.5 dB). In order to reduce losses, the input channel is made of

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circular oversized waveguides 19.6 mm in diameter. Figure 1 shows a schematic of the mixing chamber, and Figure 2 shows its general appearance. The pressure-type superconducting point contacts (STK) used in it have a resistance in the normal state of $R = 15 \dots 80$ ohms and characteristic voltage $V_c = I_c R = 300 \dots 600$ microvolts; I_c -- critical current of STK.

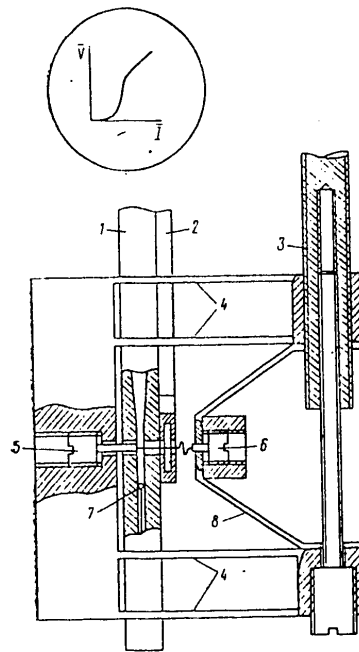


Figure 1. Schematic of the mixing chamber: 1 -- signal waveguide; 2 -- heterodyne waveguide; 3 -- differential screw; 4 -- parallelogram springs; 5 -- target; 6 -- needle holder; 7 -- piston; 8 -- V-shaped spring. The insert shows the volt-ampere characteristic of a typical superconducting point contact used in radio astronomy observations.

The needle was sharpened and the end of the target was cleaned on a special disc covered with abrasive paper No 00. The disc was rotated (≤ 10 rpm) by a micromotor. After the treatment on the abrasive paper, the electrodes were washed in oxygen-free water. The assembled mixing chamber was immediately placed in a thin-walled stainless steel housing. The housing was filled with gaseous helium, which prevented the contamination of the electrodes. The exposure of the electrodes to atmospheric air did not exceed 15 minutes. The housing protected the mixing chamber against shocks during its submerging into the Dewar vessel, and also made it possible to cool the chamber in a Dewar vessel with liquid nitrogen prior to its submerging into liquid helium.

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The contacting was done after the mixing chamber was placed in the Dewar vessel filled with liquid helium. The formation of the STK and the selection of its parameters were done with the aid of a pull rod extended to the cap of the Dewar vessel. The pull rod put in motion a pair of parallelogram springs rigidly connected with the needle. The use of the parallelogram pair ensured both a smooth contacting of the electrodes and the protection of the STK against external mechanical influences (shocks).

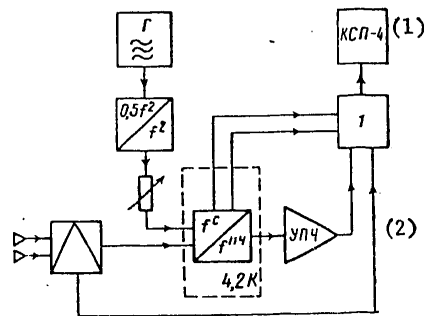


Figure 3. Block diagram of the 2-mm band radiometer: 1 -- circuit of prescribing the operating point and joining with the low-frequency part of the modulation radiometer.

Key: 1. KSP-4 recorder
2. UPCh [intermediate-frequency amplifier]

Figure 3 shows a block diagram of the radiometer. A semiconductor frequency doubler fed from a standard 4-mm generator G3-37 was used as a heterodyne. Heterodyne radiation was delivered to the STK through a separate channel. The connection between the signal channels and the heterodyne was accomplished only through the flow of current through the STK, i.e., it was reduced to a minimum. The latter reduced the depths of parasitic modulation of the heterodyne radiation at the input modulator of the radiometer [4]. When the 10-liter transportation helium Dewar vessel containing the mixing chamber was filled completely, it was possible to work continuously in the course of 48 hours. This length of operation is not at all maximal and later it will be increased by at least one order through the introduction of thin-walled waveguides of stainless steel, instead of the copper ones, in the section of the heterodyne channel, as well as by using a Dewar vessel which will make it possible to add liquid helium periodically in the process of operation.

The low-frequency part of the modulation radiometer was structurally connected with the circuit for prescribing the operating point (SZRT). The SZRT ensured both the observation of the volt-ampere characteristics (VAKh) of the STK on the oscillograph (sweep frequency 90 Hz), and their recording on a XY-recorder (sweep period from 10 seconds to 40 minutes). Moreover, the SZRT made it possible to regulate independently the amplitudes of the variable and constant components of the current flowing through the STK. The insert in Figure 1 shows the VAKh of typical STK used in radio astronomy observations. In order to suppress the modulation-demodulation [4],

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the work was done in the mode of prescribing a constant voltage on the STK, which was ensured by a circuit of low-frequency shunting of the STK 'uilt into the SZRT.

A transistor amplifier with a noise factor of ~ 3 dB and a band of amplified frequencies $f_{\text{пч}} = 200 \dots 400$ MHz was used as an UPCh. The low value of $f_{\text{пч}}$ made it possible to perform all measurements only in a two-band mode. In order to reduce the influence of the background and to suppress parasitic signals on frequencies multiple to the frequency of the signal, a resonant diaphragm similar to the one described in work [4] was installed directly before the mixer. Registration of signals was done with the aid of a KSP-4 recorder connected at the output of the synchronous detector.

The value of ΔT_{f1} was measured in the process of the laboratory tests of the radiometer. Measurements were done with the aid of two matched loads applied to the inputs of the pattern switch. One of the loads had a constant temperature of 290 K, while the temperature of the other load was changed from 290 to 77 K. The value of the signal/noise ratio measured at that time corresponded [1] to ΔT_{f1} adjusted to the input of the pattern switch, being equal to 0.2...0.3 K at a time constant of 1 sec for the overwhelming majority of the contacts. The best registered value of ΔT_{f1} was 0.12 K, which is not any inferior than the sensitivity of the best modern superheterodyne radiometers of the 2-mm band [5] which, however, use UPCh having a noise temperature which is 4...5 times lower.

Since the main contribution to the total noise of the receiver was that of the UPCh, further improvement of the sensitivity of the radiometer must be connected primarily with the use of transistor amplifiers with a lower value of the noise factor. It is feasible to reduce the ΔT_{f1} to approximately one half. It is also possible to increase somewhat the conversion factor and, accordingly, to increase the sensitivity by narrowing the band of the input filter (approximately by a factor of 1.5). A more radical increase of sensitivity is possible through the use of cooled UPCh. For example, by using a maser as an UPCh [6], it is possible to improve the ΔT_{f1} approximately to 0.02 K, which is substantially better than the ΔT_{f1} which can be obtained with mixers of other types.

2. Results of Radio Astronomy Observations

In March 1979, the radiometer was installed on a radio telescope of the millimeter wavelength band RT-25X2 [7] of the IPF of the USSR Academy of Sciences (Gor'kiy).

The matching of the radiometer with the antenna of the radiotelescope was accomplished with the aid of two parabolic-horn feeds of the 2-mm band [8] which had a horizontal diversity of 4 angular minutes. The horns ensured cosinusoidal irradiation of the parabolic reflector and reception of radiation from a source of only one polatization -- vertical. The losses in the feeds did not exceed 2 dB.

The antenna system of the RT-25X2 is periscopic with a flat reflector with whose aid the elevation guidance to the source is accomplished and which ensures (at its vertical position) the mode of monitoring and studying the directional characteristics of the antenna by the autocollimation method [7,9]. The autocollimation method in application to the RT-25X2 makes it possible to register the field distribution in the focus of the antenna and thus to measure the directional pattern of the telescope in its far zone. Figure 4 shows a horizontal section of the directional

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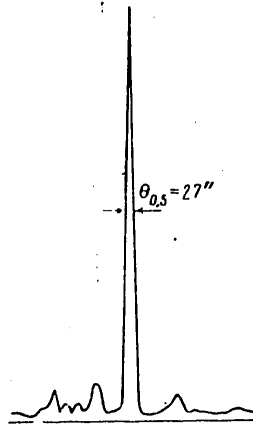


Figure 4. Horizontal section of the directional pattern of the RT-25X2 on a 3-mm wave obtained by the autocollimation method.

pattern of the RT-25X2 on a 3.1-mm wave obtained by the autocollimation method with the aid of the radiometer described above. The directional pattern of the RT-25X2 is of the beavertail-type -- narrow in the horizontal plane and wide (wider by approximately one order) in the vertical plane. On a 3.1-mm wave, the width of the principal lobe of the directional pattern at the half-power level was $\theta_{0.5h} = 27''$ in the horizontal plane and $\theta_{0.5v} = 5'$ in the vertical plane. On a 2.2-mm wave, we have respectively $\theta_{0.5h} = 20''$ and $\theta_{0.5v} = 3.5'$. The level of the side lobes in this case, according to our evaluation, was of the order of 14% of the principal with respect to the amplitude, the antenna surface utilization factor was about 30%, and the effective surface of the antenna was 15 m². The threshold signal level on the antenna was 0.3 K. Somewhat lower sensitivity in comparison to that measured under laboratory conditions can, possibly, be explained by the losses introduced by the feeds of the antenna.

During the testing of the radiometer on the RT-25X2, radio emission of the sun, moon, and the atmosphere was measured. It was impossible to fulfill the program of observations of planets and discrete sources due to bad weather conditions.

Figures 5 and 6 show central scans -- recordings of the passing of the sources of radio emission through the directional pattern of the telescope (the sun and the moon, respectively). Observations of the moon and the sun were conducted on the 27, 28, and 29 of March 1979. On March 28th, the sun and the moon had almost equal heights at the moment of culmination (difference in the heights -- 1.3 degrees), and the moments of observation differed only by 16 minutes. The phase angle of the moon was equal to 177.7 degrees; the illuminated part of the disk of the moon $K=0$ -- new moon.

In such a case, it is possible to assume that the absorption of radio emission in the earth's atmosphere on the reception beams remained practically constant at the

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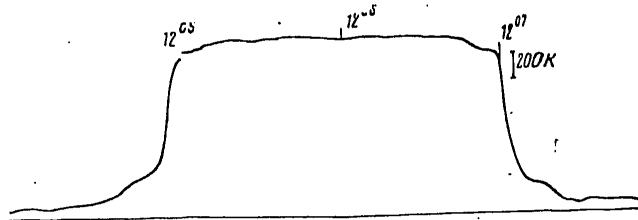


Figure 5. Central scan of the sun on a 2.2-mm wave obtained on the RT-25X2 on 28 Mar 79, $T_{ca}=1424$ K -- antenna temperature of the sun.

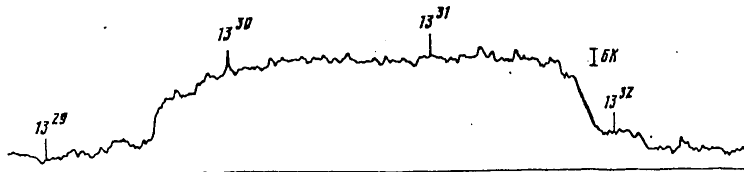


Figure 6. Central scan of the moon on a 2.2-mm wave obtained on the RT-25X2 on 29 Mar 79, $T_{\mu} = 39$ K -- antenna temperature of the moon.

moments of observation of the sun and the moon, and the directional characteristics of the antenna did not change when it was retuned from the sun to the moon. With these assumptions, let us evaluate the brightness temperature of the sun on a 2.2-mm wave by the new moon which we shall use as a radiation standard [10].

Let us calculate the temperature of the moon on a 2.2-mm wave. Let us take into consideration that the directional pattern of the RT-25X2 is of the beavertail-type and, therefore, the effect of the averaging of the moon temperature by the directional pattern [11] in our case can be considered only in one plane -- vertical, taking only the dependence of the moon temperature on the selenographic latitude. We shall take the form of the directional pattern within the limits of the disk of the moon on the basis of measurements by the autocollimation method. The brightness temperature of the moon was calculated by the relation (13) from work [12] on the basis of the available data of works [13,14] with consideration for the latitudinal changes of the reflection coefficient for vertical polarization. The table shows: values of the phase angle of the moon Φ^0 , brightness temperatures of the center of the moon $T_{\mu}(0)$, values of average brightness temperatures of the moon \bar{T}_{μ} with consideration for the directional pattern and brightness temperatures of the sun $T_c = k\bar{T}_{\mu}$, where $k = \alpha_c / \alpha_{\mu}$ -- ratio of the increment of the signal from the sun to the increment of the signal from the moon at the input of the radiometer. It can be seen that the brightness temperature of the sun on a 2.2-mm wave according to our data is $T_c = 5300 \pm 100$ K. We did not take into consideration the possible errors due to inaccurate orientation of the antenna, due to instrument fluctuations and changes of the coefficients of absorption in the atmosphere, as well as due to changes in the characteristics of the antenna during its retuning.

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Date	ϕ^*	$T_{\text{д}}^{(0)}, \text{K}$	$\bar{T}_{\text{д}}, \text{K}$	k	$T_{\text{с}}, \text{K}$
27/11-79 r.	164,5	153	146	36,7	5366
28/11-79 r.	177,7	139	143	36,8	5268
29/11-79 r.	167,8	152	146	36,2	5282

The result obtained by us differs from the accepted value of the brightness temperature of the sun in the 2-mm wavelength band equal to 5600 K. It is impossible to explain this difference, for example, by the absorption of radiation proceeding from the lower layers of the chromosphere of the sun or by high and cold formations (spicula), since the level of the sun in the scans was practically over the entire disk (within the limits of 0.8 of the sun's diameter). It is evident that it is necessary to select the parameters determining the brightness temperature of the moon more correctly.

The authors are grateful to the members of the IPF of the USSR Academy of Sciences who participated in the preparations and observations on the RT-25X2, to V. D. Krotikov for the fruitful discussions of the results of observations, as well as to the members of the Radio Engineering Institute-7.5/250 of the MVTU [Moscow Higher Technical School] imeni N. E. Bauman, where preliminary tests of the first variant of the radiometer were conducted.

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MICROELECTRONICS

INSTRUMENTS OF FUNCTIONAL SEMICONDUCTOR ELECTRONICS

Moscow NOVOYE V ZHIZNI, NAUKE, TEKHNIKE. SERIYA "RADIOELEKTRONIKA I SVYAZ": PRI-BORY FUNKSIONAL'NOY POLUPROVODNIKOVY MIKROELEKTRONIKI in Russian No 2, Feb 81 (signed to press 13 Jan 81) pp 2-5

[Annotation, foreword by L. D. Bakhrakh, corresponding member of the USSR Academy of Sciences, and table of contents from book "News in Life, Science and Technology. Series 'Radioelectronics and Communications': Instruments of Functional Semiconductor Electronics", by Dmitriy Vasil'yevich Igumnov, candidate of technical sciences, and Galina Petrovna Kostyunina, candidate of physical and mathematical sciences, Izdatel'stvo "Znaniye", 38,180 copies, 64 pages]

[Text] This pamphlet treats the most developed directions of modern functional semiconductor microelectronics and the effects on which they are based. More attention is given to instruments with the MOP [metal-oxide-semiconductor] structure.

Foreword

Microelectronic products greatly improve technical and economic indexes of electronic equipment and open up a number of new possibilities. Solid integrated circuits whose main elements are designed as transistor structures of various properties and types have become the most popular. However, further development of transistor microelectronics has a theoretical limit due to design complexity, decrease in the reliability, power consumption, etc.

These limitations can be overcome only by changing over to the development of new directions in microelectronics. The main tendency of such development amounts to functional consolidation of structural devices. This consolidation is feasible with the use of physical phenomena which make it possible, with the aid of simple unseparable structures, to perform the functions which are usually performed with the aid of a complex multielement circuit or device. Implementation of this principle is due to the appearance of new types of devices which are usually called functional and, consequently, a new stage in the development of microelectronics: functional microelectronics.

In functional microelectronics, the carrier of information is a multivariant signal whose parameters are controlled by the dynamic nonuniformities of the medium occurring under the effect of the control signal. In functional devices, it is often impossible to isolate the areas performing definite specialized functions which are

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capable of processing multivariant functions. The use of functional devices increases considerably the productivity of information processing systems at prescribed dimensions and power consumption, which is equivalent to a sharp increase in the degree of integration in comparison with the classical integrated circuits.

In functional microelectronics, the element itself is sometimes multifunctional. In addition to its main functions, it is capable of performing other functions.

An ideal functional device capable of performing the function of some device as a whole.

Functional microelectronics is characterized by the use of a large number of different phenomena. Apart from purely electrical circuits, it utilizes optical, acoustic, magnetic, chemical and other phenomena in solids.

This pamphlet describes instruments of functional semiconductor microelectronics.

Along with the instruments of functional microelectronics which have been developed, the authors discuss the ways of designing new instruments, as well as the tendencies of the development of functional microelectronics. Therefore, various directions of functional microelectronics are described in the pamphlet at different lengths.

Contents	Page
Foreword	3
Instruments Based on the Gunn Effect	5
Instruments with the MOP Structure	9
Other Instruments of Functional Microelectronics	29
Instruments of Acoustic Electronics	43
Instruments of Magnetoelectronics	55
Instruments of Cryoelectronics	58
Instruments of Chemotronics	60
About Biotronics	62
Conclusion	63
Bibliography	64

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PUBLICATIONS, INCLUDING
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UDC 621.314.214

BOOK DISCUSSES DESIGN AND OPTIMIZATION CRITERIA FOR SEMICONDUCTOR DEVICES

Kiev OPTIMIZATSIYA POLUPROVODNIKOVYKH USTROYSTV ENERGETICHESKOY ELEKTRONIKI
in Russian 1980 (signed to press 29 Sep 80) pp 2, 145

[Annotation and table of contents from book "Optimization of Semiconductor Devices
for Power Electronics", edited by V. Ya. Pekurovskiy and L. D. Prokopenko,
Izdatel'stvo "Naukova dumka", 800 copies, 145 pp]

[Text] The collection contains articles devoted to questions of the optimization
and design of semiconductor devices intended for transformation of the parameters
of electromagnetic energy: frequency converters, rectifiers, voltage regulators
and stabilizers and secondary power sources.

It is intended for scientific and engineering and technical personnel specializing
in the area of converter technology and automation equipment.

CONTENTS

	Page
Chizhenko, I. M., Kurilo, I. A., Yakimov, O. S. Transient conditions of a current inverter given abruptly changing capacitive load	3
Tonkal', V. Ye., Grechko, E. N., Alymov, O. P. Synthesis of an automatic control device for the digital control system of an autonomous transistorized inverter	10
Sen'ko, V. I., Rudenko, V. F., Skobchenko, V. M., Smirnov, V. S. On the optimum synthesizing of AM voltage	20
Perkhach, V. S. Analysis and optimization of the conditions of electric power systems with rectifier devices using digital modeling methods	31
Lipkovskiy, K. A., Sidorenko, Yu. V. Improving the efficiency with which switch elements are used in alternating voltage regulator actuators	40

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Budennyy, A. V., Sen'ko, V. I. Ways of improving the weight-size characteristics of converters with summation in the common circuit	46
Pirnyak, G. G. Peculiarities of using d.c. thyristor drive for contactless electric locomotives	53
Bezgachin, N. I. Analysis of the frequency characteristics of the overall power of the reactive elements of converter resonance circuits	60
Molodyko, B. K., Sinitsina, N. M. Selection of transistors and their operating conditions in a synchronous rectifier	69
Mezenina, N. S. Ways to limit unidirectional saturation of a transistorized converter transformer	77
Alimov, O. P. Control system for an autonomous transistorized inverter with an output transformer	85
Verteletskiy, D. S. Device for tuning and control of vibration machines with electromagnetic drive	93
Grechko, E. N., Adamishin, M. M. Regulation of the line voltage fed to a synchronous motor on the basis of ShIM [pulse-duration modulation] of alternating voltage curves	97
Mel'nichuk, L. P., Dykhnenko, Yu. I., Novosel'tsev, A. V., Griban', N. G. A digital control system for a frequency converter with an intermediate alternating current link	104
Anisimov, Ya. F., Petrov, V. M., Skorokhodov, V. A., Simonyan, S. T. Low-frequency fluctuations in the output voltage of static converters	111
Mogilevskiy, G. V., Dikan', S. V. Induced switching units of a thyristor with current switch-over into a high-frequency thyristor	117
Shipillo, V. P., Chikotilo, I. I. Dynamics of a closed automatic control system with ShIPPT [expansion unknown] under intermittent current conditions	124
Goncharov, Yu. P. Methods for studying steady-state and transient processes in rectifier-type switches	127
P'yanykh, B. Ye., Chekhet, E. M. Some power characteristics of transistorized frequency converters with one-time modulation	130
Lipkovskiy, K. A. Three-phase transformer-thyristor regulators and discrete action stabilizers	139

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A CONCISE DICTIONARY OF RADIO-ELECTRONICS

Moscow KRATKIY SLOVAR' PO RADIOELEKTRONIKE in Russian 1980 (signed to press 19 Feb 80) pp 2-3

[Annotation and foreword from book "A Concise Dictionary of Radio-Electronics", by Anatoliy Pavlovich Verzhikovskiy (deceased), Nikolay Vladimirovich Gabis (deceased), Nikolay Mikhaylovich Kitayev, Ivan Ignat'yevich Tynyankin and Vladilen Grigor'yevich Grigor'yants, Voenizdat, 25,000 copies 512 pages]

[Text] This edition of "A Concise Dictionary of Radio-Electronics" is a second revised and enlarged edition of the dictionary published in 1964.

The dictionary includes about 4,000 terms relating to radar, radio communications, radio navigation, television, remote-control, automation, radiometeorology, hydro-acoustics, infrared technology, electronics, computer engineering, radiometrics, antenna systems, radio camouflage and electronic countermeasures; in addition, the dictionary lists the names of certain foreign radio-electronic systems and explains their designations.

The terms explained in the dictionary are arranged in alphabetical order and are set in bold-face capital letters. If a term explained in an entry is repeated, the term is designated by the initial letters of each of the constituent words in the term, each followed by a period. Certain of the terms explained are followed by their accepted abbreviations and explanations in parentheses. Explanations which are necessary for revealing the essence of the term and which are discussed in other articles are set off in italics. A system of symbols and abbreviations has been extensively employed in the book in order to reduce the length of the entries.

Open-source domestic as well as foreign (American, British, French and German) books, technical and military journals, reference books, official publications and other types of literature were used in compiling the book.

This dictionary is intended for a broad circle of readers concerned with questions of radio-electronics.

We request that all comments and wishes regarding the content and design of this dictionary be sent to: 103160, Moscow, K-160, Voennoye izdatel'stvo.

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9512

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32

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UDC 621.396:519.2-526

CONTROL OF TERRESTRIAL ANTENNAS AND RADIOTELESCOPES

Moscow UPRAVLENIYE NAZEMNYMI ANTENNAMI I RADIOTELESKOPAMI in Russian 1980
(signed to press 16 Jul 80) pp 2, 278-279

[Annotation and table of contents from the book "The Control of Terrestrial Antennas and Radiotelescopes" by Pavel Vasil'yevich Belyanskiy and Boris Georgiyevich Sergeyev, Izdatel'stvo "Sovetskoye radio", 5000 copies, 280 pages]

[Text] This work treats problems in the design of modern systems for the automatic control of ground-based antennas within basic operating regimes taking into account control and disturbance effects, as well as the dynamic characteristics of an antenna employed as a control device.

It is intended for engineers, scientific researchers, and designers working in the field of antenna design and application, radiotelescope and optical telescope design, space communications, and radioastronomy, as well as for graduate and post-graduate students.

Table of Contents

Introduction	3
Chapter 1	
The Kinematics of Tracking Objects in Space Using Antenna Arrays	11
1.1. The celestial sphere and coordinate systems	12
1.2. Variation in the coordinates of space objects	13
1.3. Functional diagrams of support/steering devices and coordinates for antenna arrays	16
1.4. Kinematics of tracking distant space objects	20
1.5. Breaks in communication and the strategy of singular point passage	28
1.6. Kinematics of tracking nearby space objects	31
Chapter 2	
The Principles of Design and Basic Features of the Automatic Control System for Terrestrial Antenna Arrays	44

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2.1. Basic requirements for SAU[automatic control systems] for ground-based antenna arrays	44
2.2. Basic operating regimes of SAU for ground-based arrays	50
2.3. Design and mathematical description of an SAU for ground-based antenna arrays	59
2.4. Design principles and basic features of SAU for large antennas and radiotelescopes	70
Chapter 3	
Equipment for Programmed Guidance in Automatic Control Systems for Ground-Based Antenna Arrays	87
3.1. Design of equipment for programmed control	87
3.2. Schematic implementation of digital linear interpolators	94
3.3. Digital integrators in SAU for ground-based antenna arrays	100
3.4. Use of digital computers in SAU for ground-based arrays	106
Chapter 4	
Guidance Mechanisms and the Structural Design of Support/Steering Devices for Terrestrial Antenna Arrays Used as Control Devices	116
4.1. Development of metallic designs and contours of ground-based antenna arrays	116
4.2. Mathematical models of ground-based antenna arrays as control devices	119
4.3. Methods for the construction of dynamic models of ground-based antenna arrays as control devices	122
4.4. The effect of external factors on ground-based antenna arrays	136
4.5. Dynamic characteristics of ground-based antenna arrays	143
Chapter 5	
Power-Actuated Servo Drive of Ground-Based Antenna Arrays	154
5.1. Basic requirements of SSP[powered servo drive] in ground-based antenna arrays	155
5.2. Design and characteristics of SSP in ground-based arrays	158
5.3. Selection of the primary elements and parameters of SSP in ground-based antenna arrays	170
5.4. Special factors in the design of SSP for ground-based arrays	180
Chapter 6	
Precise Guidance of Ground-Based Antenna Arrays	
6.1. Problems in achieving precise guidance and quality criteria in the operation of SAU of ground-based antenna arrays	187
6.2. Error classification and total error in the guidance of ground-based antenna arrays	189
6.3. Effect of deformation of the antenna reflecting system on its electrical characteristics and methods of compensation for such deformation	197
6.4. Dynamic error in the processing of useful input signal in SAU of ground-based antenna arrays	205

FOR OFFICIAL USE ONLY

6.5. Random dynamic guidance errors in SAU of ground-based arrays	207
6.6. Distribution of guidance errors in the passband and selection of the optimal bandwidth for SAU of ground-based arrays	213
6.7. Synthesis of SAU of ground-based antenna arrays	219
Chapter 7	
Increasing the Dynamic Accuracy of SAU in Ground-Based Antenna Arrays	225
7.1. Application of invariance principles to SAU of ground-based antenna arrays	225
7.2. Optimal SAU of ground-based antenna arrays	241
7.3. Design problems in the optimal SAU of ground-based arrays	256
Bibliography	268
Subject index	275

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9481

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COMPONENT STRUCTURE IN LARGE-SCALE INTEGRATED CIRCUITS

Novosibirsk STRUKTURA KOMPONENTOV BIS in Russian 1980 (signed to press 1 Oct 80) pp 2, 255-6

[Annotation and table of contents from the book "Component Structure in LIC", by Feliks Leonidovich Edel'man, Izdatel'stvo "Nauka", 1850 copies, 256 pages]

[Text] This book examines the structural features of components in LIC (large-scale integrated circuits)--the fundamental building blocks of computer technology: silicon dioxide films, silicon nitride films, and polycrystalline silicon films. Based in large measure on original research of the author, the book presents a picture of the close relationship of component design improvement to LIC technologies, of phase changes in films, and of contact interactions of materials. Included is a special discussion of the capabilities of current methods of structural analysis used for component research and structural correlation to LIC technology, as well as for description of semiconductor surfaces. Problems of LIC materials technology are examined with a view to current and future generations.

This monograph is intended for use by scientific and technical engineering personnel specializing in the field of semiconductor materials, instruments, and microcircuits.

Table of Contents

Foreword	3
Chapter 1. Silicon Dioxide Films on Silicon	6
1.1. Use of silicon dioxide	6
1.2. Crystalline silicon dioxide	8
1.3. Films of amorphous silica(short-range order)	12
1.4. Oxidation of silicon. Silicon-oxide interface	18
1.5. Damage to silicon occurring during oxidation,-- --stacking faults	25
1.6. Crystallization of silicon dioxide films on silicon	29
1.7. Phenomena associated with crystallization of silicon dioxide films on silicon	51

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1.8.	Effect of the implantation of ions on the structure of thermic silicon dioxide	67
1.9.	Mechanical stresses in MOS and MNOS structures on silicon	69
1.10.	Interaction of silicon dioxide with metal films used for dividing current in LIC	75
1.11.	Structure of aluminum oxide and aluminosilicate films used in LIC	84
1.12.	Hillocks in LIC aluminum film conductors	95
	Bibliography	96
Chapter 2.	Silicon Nitride Films	107
2.1.	General properties. Synthesis	107
2.2.	Stoichiometric silicon nitride	109
2.3.	Non-stoichiometric silicon nitride	120
2.4.	Oxidation of silicon nitride and technical defects	132
2.5.	Silicon nitride films and broadband semiconductors developed on the surface of monocrystalline silicon by the ionic implantation method	143
	Bibliography	145
Chapter 3.	PSF--Polycrystalline Silicon Films	149
3.1.	General properties and application	149
3.2.	Pyrolysis of silane	156
3.3.	Nucleation of PSF. Textures. Recrystallization	164
3.4.	Low-temperature amorphous silicon films(ASF)	177
3.5.	Electrical conductivity of PSF. Alloying	182
3.6.	Etching	187
3.7.	Morphological defects	188
	Bibliography	190
Chapter 4.	Dynamics of Defects in Semiconductors and Some Problems of the Planar Technology of A_3B_5 and Germanium	195
4.1.	Microhardness of semiconductors and the process of encription	195
4.2.	The nature of dislocation in silicon at high temperature and under constantly applied stress(creep)	196
4.3.	Experiments related to internal friction in germanium and silicon	198
4.4.	Electrophysical properties of semiconductors with high dislocation density(indium antimonide)	199
4.5.	An unsuccessful attempt to introduce a uniform system of dislocation into a semiconductor	200
4.6.	Observation of polarized facets in A_3B_5 semiconductors and analysis of anodic oxide on indium antimonide	201
4.7.	Structure of oxide films on gallium arsenide and germanium	218
	Bibliography	219

FOR OFFICIAL USE ONLY

Chapter 5. Methods of Structural Analysis of LIC Components	222
5.1. Traditional methods	222
5.2. Newer methods	226
5.3. Advanced methods	232
Bibliography	235
Chapter 6. Ten Interviews with the Inventors of LIC	236
Conclusion	251

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9481
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CURRENT STABILIZATION METHODS AND EQUIPMENT

Kiev METODY I TEKHNICHESKIYE SREDSTVA STABILIZATSII TOKA in Russian 1980 (signed to press 8 Aug 80) pp 157-64

[Abstracts of articles from book "Current Stabilization Methods and Equipment", edited by A. I. Kuz'menko, Izdatel'stvo "Naukova dumka", 950 copies, 164 pages]

UDC 621.372.061

A METHOD OF REALIZING MULTIPHASE ELECTRIC CIRCUITS WITH GIVEN CHARACTERISTICS

[Abstract of article by Volkov, I. V., Gubarevich, V. N., Isakov, V. N., and Kaban, V. P.]

[Text] A method of constructing multiphase electric circuits with given characteristics is described on the basis of the utilization of simple active four-terminal structures. The article presents the future prospects for the application of this method in the construction and analysis of multiphase circuits of various functional purposes. 2 illustrations, 2 tables, 2 titles in bibliography.

UDC 621.311.6

A PASSIVE FOUR-TERMINAL TRANSFER NETWORK OPERATING WITH A CONSTANT INPUT CURRENT

[Abstract of article by Volkov, I. V., Zakrevskiy, S. I., and Basan'ko, Yu. V.]

[Text] This article examines the operation of a passive four-terminal transfer network with a constant-value input current when the load impedance varies within a broad range. It also examines the selection specifications for the circuit elements of particular devices and cites the relative and external characteristics of these devices. 3 illustrations, 4 titles in bibliography.

UDC 621.314.2.015.4

A GRAPHIC METHOD OF DETERMINING CURRENTS CONSUMED BY A THREE-PHASE INDUCTIVE-CAPACITIVE CONVERTER

[Abstract of article by Obukh, A. I.]

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[Text] The article describes two graphic methods of determining linear currents consumed by a three-phase inductive-capacitive converter (IEP) with any combination of load impedances Z_L . The first method is based upon the use of a topographic diagram (nomograph) of voltages in the IEP, constructed beforehand using given values of Z_L . The second method--a method of current polar nomographs--makes it possible to determine quickly the current values directly from the parameters Z_L . 3 illustrations, 5 titles in bibliography.

UDC 621.372.061

ANALYSIS OF A THREE-PHASE BRIDGE-TYPE INDUCTIVE-CAPACITIVE CONVERTER BASED ON THE TRANSFORMATION RATIO

[Abstract of article by Zavarikhin, V. A., and Kosovskaya, L. V.]

[Text] This article examines a bridge-type inductive-capacitive converter with a current shift angle of 240° in each rod of the choke and with opposing coils, based on the transformation ratio of the choke. The basic relationships between currents and the installed capacities of the converter elements are obtained. It is shown that the minimum of the specific reactive capacities shifts to the right with an increase in the transformation ratio. 2 illustrations, 4 titles in bibliography.

UDC 621.372.161

A REFINED ANALYSIS OF THE PROCESSES IN A CHARGING CIRCUIT WITH A SINGLE-PHASE L-SECTION INDUCTIVE-CAPACITIVE CONVERTER

[Abstract of article by Pentegov, I. V.]

[Text] This article examines the processes involved in charging a bank of capacitors connected to a single-phase L-section inductive-capacitive converter through a rectifier bridge when it is assumed that an analysis of the processes in each of the half-cycles can be accomplished by a method of calculating on the counterelectromotive force.

It is shown that the charging current in this case is not constant and that it decreases with respect to an increase in the voltage on the capacitors. Formulas are presented which make it possible to calculate the parameters of the charging process. 4 illustrations, 3 titles in bibliography.

UDC 621.314.5

DETERMINING TRANSFORMATION RATIOS OF A RECTIFIER SUPPLIED BY A SOURCE OF STABILIZED CURRENT WHEN CHARGING A CAPACITIVE STORAGE CIRCUIT

[Abstract of article by Spirin, V. M., and Kurach, A. M.]

[Text] The article gives a determination of the transformation ratios of a rectifier supplied by a source of stabilized current based on the inductive-capacitive transformation of the voltage surge into a current source during the changing of a capacitive storage circuit of a pulse current generator (GIT).

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The determination of the ratios is carried out with the use of the theory of experimental design, in particular, the synthesis of a second-order orthogonal design. The ratios are represented by polynomials for the encoded and natural values of the selected factors. The information in the article makes it possible to determine the transformation ratios of rectifiers for pulse current generators of a parametric number of electrohydraulic installations. The data can also be used to design pulse current generators. 2 illustrations, 2 tables, 4 titles in bibliography.

UDC 621.314.5

OPERATION OF A THREE-PHASE BRIDGE RECTIFIER FROM THE CURRENT SOURCE TO THE CAPACITIVE STORAGE CIRCUIT

[Abstract of article by Spirin, V. M.]

[Text] This article analytically determines the transformation ratios of a rectifier fed from a source of stabilized current and based on the inductive-capacitive transformation of a voltage source to a current source when there is a slow charge of a capacitive storage circuit of a pulse current generator (GIT). The article compares the ratios obtained with analogous ratios when the rectifier is supplied from a voltage source and a source of stabilized current for various types of loads. The data obtained make it possible to determine the transformation ratios of rectifiers for GIT charging devices and can be used to design pulse current generators. 2 illustrations, 1 table, 7 titles in bibliography.

UDC 621.3:621.791.75

A STATIC CONVERTER FOR CHARGING A RESERVOIR CAPACITOR WITH A MINIMAL EFFECT ON THE SUPPLY CIRCUIT

[Abstract of article by Kofman, D. B., Lomonosov, L. Ye., and Chorba, V. R.]

[Text] This article examines a new structure for a static converter operating in a constant power-consumption mode, used for charging a reservoir capacitor to a voltage considerably in excess of the voltage in the supply circuit. The article presents the relationships between the parameters of the converter and its analog, with the help of which one can determine the voltages and currents necessary for the efficient selection of the circuit elements. 2 illustrations, 3 titles in bibliography.

UDC 621.313.67.001.5

AN ANALYSIS OF THE STEADY-STATE BEHAVIOR OF A CONVERTER-RECTIFIER-MOTOR SYSTEM

[Abstract of article by Il'inskiy, N. F.]

[Text] The physical processes in a system composed of a three-phase inductive-capacitive converter, a rectifier and a DC motor are examined. Relations are obtained for determining the circuit's power factor and the distortion factor. Experimental results confirming the theoretical conclusions are cited. 10 illustrations, 2 titles in bibliography.

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UDC 621.314.562

DIRECT-CURRENT ELECTRIC DRIVE USED IN POLYMER FILM WINDING MACHINES SUPPLIED BY THE ARMATURE CIRCUIT OF A RECTIFIER-CAPACITIVE CURRENT SOURCE

[Abstract of article by Arkushin, V. P., and Osetskiy, Yu. M.]

[Text] This article presents a description of an electric drive for winding machines. An equation is derived for the external characteristic of a single-phase current source for uninterrupted current from the armature circuit. Expressions are also derived for selecting the power of the capacitor bank. The accuracy of the calculated relations is checked by experiment. 3 illustrations, 3 titles in bibliography.

UDC 621.314.572

STABILIZING CHARACTERISTICS OF FREQUENCY-DEPENDENT FOUR-TERMINAL NETWORKS IN THE CURRENT-STABILIZING MODE

[Abstract of article by Starodumov, Yu. I.]

[Text] This article examines the operation of current-stabilizing four-terminal networks with frequency-dependent parameters in circuits with loads that vary in value when the frequency of the supply line changes. An expression in a general form is obtained for the stabilization factor with respect to the load of a four-terminal network, based on the frequency variation. The article points out the feasibility of calculating the influence of a matching transformer upon the stabilizing characteristics of a four-terminal network. 3 titles in bibliography.

UDC 621.372.061

THE RIT-1 REGULATED STABILIZED CURRENT SOURCE FOR POWER SUPPLY TO LASERS

[Abstract of article by Zakrevskiy, S. I., and Gorbachev, M. N.]

[Text] The principle of circuit construction of an RIT-1 regulated stabilized current source is presented. Its description is given and its basic specifications are cited. 1 illustration, 4 titles in bibliography.

UDC 621.372.54.2.001.24

INSTALLED CAPACITY OF CAPACITORS AND CHOKES IN CIRCUITS WITH NONSINUSOIDAL VOLTAGE AND CURRENT

[Abstract of article by Nikitin, V. B., and Bezgachin, N. I.]

[Text] Several determinations of the installed capacity of reactive elements in circuits with nonsinusoidal voltage and current are examined. It is shown that the summation of reactive capacities of individual harmonics must of necessity be carried out using factors inversely proportional to the frequency of these harmonics when determining the installed capacity of ideal capacitors and chokes. A frequency-

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dependent factor is introduced in order to calculate the frequency dependence of the specific reserve power of real capacitors and chokes. This makes it possible to utilize more reasonably the concept of installed capacity in calculating the weight and dimension indicators of reactive elements. 1 illustration, 1 table, 5 titles in bibliography.

UDC 621.314.621

A STUDY OF CASCADING OF ACTIVE POWER FILTERS WITH SERIES COMPENSATION

[Abstract of article by Ryaben'kiy, V. M.]

[Text] This article examines a filter device with a series-connected L-section LC-filter and two active filters with series compensation. Studies are carried out on the structure of the device and its efficiency. Recommendations are developed for its design. 2 illustrations, 2 titles in bibliography.

UDC 621.314.621

A HARMONIC ANALYSIS OF RECTIFIED VOLTAGE OF STABILIZERS WITH INVERTER REGULATION

[Abstract of article by Ryaben'kiy, V. M., Simonyan, S. T., and Shvets, E. A.]

[Text] This article examines a harmonic analysis of the rectified voltage of a converter with voltage stabilization on the alternating-current side, achieved with an inverter. An analysis of the level of harmonics is done for the instances of symmetrical and nonsymmetrical voltages in the supply line. 3 illustrations, 3 titles in bibliography.

UDC 621.3:621.791.75

AN INVESTIGATION INTO THE CHARACTERISTICS OF ALTERNATING-CURRENT SOURCES WITH INDUCTIVE POWER-STORAGE CIRCUITS WHEN THERE IS PARAMETRIC STABILIZATION OF THE WELDING PROCESS

[Abstract of article by Legostayev, V. A., Pentegov, I. V., Stenkovskiy, Ye. P., and Chayun, A. G.]

[Text] A comparison is made of two types of AC welding sources with inductive power-storage circuits regarding the power parameters of their charging devices, their efficiency and the power liberated in the arc interval. In order to facilitate the development of technical recommendations, a system of dimensionless criteria are proposed for the welding current. 4 illustrations, 5 titles in bibliography.

UDC 621.372.061

THE RIT-300 WIDE-RANGE REGULATED STABILIZED-CURRENT SOURCE FOR SUPPLYING A SUPER-CONDUCTING SOLENOID

[Abstract of article by Gorbachev, M. N., Stepanov, P. P., and Novobranets, V. I.]

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[Text] Questions are examined regarding the construction of a wide-range regulated current source for superconducting magnetic systems. A description is provided for the principal circuit of the source. Its features and basic specifications are cited. 2 illustrations, 5 titles in bibliography.

UDC 621.378.325;625.311.6

INSTALLATION OF THE 'KVANT-1GM' FOR HEAT-TREATING MATERIALS BY LASER

[Abstract of article by Chuchumayev, Ye. F., Filippov, V. K., and Bacan'ko, Yu. V.]

[Text] An industrial installation for laser heat-treating materials is described. The operational advantages of the "Kvant-16M" installation are shown in comparison to existing units. 1 illustration.

UDC 631.371:621.311.24

OPTIMIZING THE PERFORMANCE OF WIND-POWER INSTALLATIONS

[Abstract of article by Volkov, I. V., and Muzychenko, A. D.]

[Text] A block diagram is proposed for a wind-power electric station. When the wind velocity varies, the layout provides the maximum possible output of power corresponding to the velocity. 13 titles in bibliography.

UDC 621.311.1+621.3.015+621.3.018.3

RESONANCE PHENOMENA IN THREE-PHASE ELECTRIC CIRCUITS WITH NONSYMMETRICAL REACTIVE ELEMENTS

[Abstract of article by Kuznetsov, V. G., and Danilyuk, V. B.]

[Text] The article deduces and analyzes expressions for resonance frequencies in electric circuits when symmetrical devices are connected. It is shown that a device with nonsymmetrical reactive elements can cause current resonance in the circuit at two frequencies. 1 illustration, 1 table, 3 titles in bibliography.

UDC 621.372.061

REGULATION OF OUTPUT CURRENT IN COMBINED SYSTEMS WITH INDUCTIVE-CAPACITIVE CONVERTERS

[Abstract of article by Lipkovskiy, K. A., and Aleksandrov, M. M.]

[Text] A comparison is made of various methods of regulating the current of a load supplied by a combined system with an inductive-capacitive converter. Requirements are determined for changing the parameters of additional electromotive force from the standpoint of the criterion for the maximum installed capacity of the elements. 4 illustrations, 2 titles in bibliography.

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UDC 621.377.534.032.43

CIRCUITS FOR TRIGGERING LOW-POWER GAS LASERS WITH A PARAMETRIC DEFLECTION

[Abstract of article by Dovgalevskiy, A. Yu., Sapon, O. P., and Starodumov, Yu. I.]

[Text] A comparison is made of various circuits for triggering low-power gas lasers with a parametric deflection. A circuit is suggested for separate firing of dual-mode lasers. This arrangement possesses high economy and reliability of triggering both channels. 5 illustrations, 3 titles in bibliography.

UDC 621.372.061

NONRESONANT TUNING OF REACTIVE ELEMENTS AS A MEANS OF REGULATING THE LOAD CURRENT OF INDUCTIVE-CAPACITIVE CONVERTERS

[Abstract of article by Gubarevich, V. N., and Aleksandrov, M. M.]

[Text] This article presents results of an investigation into the nonresonant operation of variously structured inductive-capacitive converters. It points out the expedience of utilizing the nonresonant method of regulating the output current of an inductive-capacitive converter. 8 illustrations, 3 titles in bibliography.

UDC 621.372.54.061

DUALITY OF AC POWER ELECTRIC FILTERS

[Abstract of article by Nikitin, V.B.]

[Text] A comparative analysis of two classes of filters is presented: filters designed to attenuate the upper harmonic components of alternating voltage used at the output of static voltage converters, and filters designed to protect an electric circuit from the intrusion of upper harmonics of the current which appear when nonlinear loads are supplied from this circuit. It is shown that the use of the principle of duality makes it possible to apply methods developed for one class of filters in order to analyze and synthesize the other class of filters. 4 illustrations, 5 titles in bibliography.

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9512
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UDC 621.391.002.72:658.284(075)

INSTALLATION AND SET-UP OF INDUSTRIAL ELECTRICAL COMMUNICATION AND ALARM SYSTEMS

Moscow MONTAZH I NALADKA SISTEM PROIZVODSTVENNOY ELEKTRICHESKOY SVYAZI I SIGNALIZATSII in Russian 1980 (signed to press 10 Jun 80) pp 2, 278-80

[Annotation and table of contents from book "Installation and Set-up of Industrial Electrical Communication and Alarm Systems", by Sergey Vladimirovich Koshelev, Grigoriy Abramovich Klauz and Viktor Vasil'yevich Gvozdevskiy, Izdatel'stvo "Svyaz'", 23,000 copies, 280 pages]

[Text] This book examines the installation and set-up of systems of electrical communication, alarms, industrial (closed-circuit) television and electric clocks used at establishments and enterprises in various sectors of the economy. Questions regarding the organization of and preparation for installation and adjustment operations, safety methods, industrial hygiene as well as fire prevention are covered.

This book is intended for students at technical schools. It can be useful for students at vocational and technical schools as well as for skilled workers, team foremen and riggers concerned with the installation and set-up of industrial communication systems.

Contents	Page
Foreword	3
Introduction	4
Chapter 1. Organization and Preparation of Production	6
1.1. General principles	6
1.2. Preparatory work	9
1.3. Inspection of the premises	10
1.4. Unpacking and inspecting equipment, materials and cables	11
1.5. The manufacturing plan	13
1.6. Organizing work at the installation site	16
Review questions	18
Chapter 2. Network Planning and Administration	18
2.1. System of network planning	18
2.2. Linear production schedule	21
2.3. Network schedule	22

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2.4. Work and resources check-off card	24
2.5. Principal responsibilities of officials	26
2.6. Responsibilities of customers and contract organizations	29
Review questions	31
Chapter 3. Safety Methods and Fire-Prevention	32
3.1. General principles	32
3.2. Loading, transporting, unloading and hoisting of equipment	34
3.3. Working at heights	35
3.4. Working with portable electric power tools and accessories	36
3.5. Working in battery rooms	38
3.6. Motor transport of personnel	39
3.7. Fire-prevention	41
Review questions	42
Chapter 4. Installing Metalwork and Machinery in the Automatic-Equipment and Terminal Rooms of an Automatic Telephone Exchange	42
4.1. Laying out the premises to accomodate the installation equipment	42
4.2. Methods of securing metalwork and equipment	45
4.3. Assembling the metalwork in the automatic-equipment room	45
4.4. Assembling metalwork in the terminal room	69
4.5. Checking the accuracy of installation of metalwork and equipment	70
4.6. Installing the current-distribution network	71
4.7. Installing overhead row lights and work lights	78
4.8. Installing grounding equipment	80
Review questions	82
Chapter 5. Cable-Installation Work	83
5.1. General principles	83
5.2. Laying cables in open trunkline and row conduits	84
5.3. Particular features of laying cables on individual segments of the row conduits	86
5.4. Particular features of laying cables when installing the K-100/2000 automatic telephone exchange	88
5.5. Laying cables in floor and underground conduits	89
5.6. Open laying of cables along walls	90
Review questions	91
Chapter 6. Installing Rack Equipment	91
6.1. General principles	91
6.2. Templates for fanning cable	94
6.3. Fanning cables	97
6.4. Fanning shielded cables	101
6.5. Connecting the cable strands	102
6.6. Testing the cable strands for continuity	110
6.7. Particular features of installing the cable at various types of automatic telephone exchanges	112
6.8. Installing cable for auxiliary equipment and test sets	131
6.9. Testing installation quality	132
Review questions	133

FOR OFFICIAL USE ONLY

FOR OFFICIAL USE ONLY

Chapter 7. Installing Power-Plants	134
7.1. General principles	134
7.2. Installing and checking rectifier units and switchboards	136
7.3. Installing acid storage batteries	140
7.4. Forming and screening storage batteries	148
7.5. Particular features of installing alkaline cells	150
Test questions	152
 Chapter 8. Adjusting Automatic Telephone Exchange Equipment	 152
8.1. General principles	152
8.2. Adjusting the main electromagnetic switching devices	153
8.3. Adjustment sequence for an automatic telephone exchange	164
8.4. Checking the quality of adjustment work	176
8.5. Preparing equipment for commissioning and accepting it for operation	177
Review questions	179
 Chapter 9. Installing and Adjusting Central Control Communication and Master Station Service	 180
9.1. General principles	180
9.2. Accomodations for and installation of equipment	181
9.3. Laying cables and wires	185
9.4. Installation requirements in explosion and fire-hazard areas	196
9.5. Setting up equipment and turning it over for operation	197
Review questions	198
 Chapter 10. Installing and Setting Up Industrial Loudspeaker Communica- tions Apparatus	 199
10.1. General principles	199
10.2. Setting up and securing equipment on various types of structural foundations	199
10.3. Interconnecting equipment components	205
10.4. Setting up equipment and turning it over for operation	208
Review questions	210
 Chapter 11. Installing and Setting Up Industrial Television Systems	 210
11.1. General principles	210
11.2. Preparing and organizing work on installing industrial television equipment	213
11.3. Installing coaxial cables	213
11.4. Setting up and securing industrial television equipment	219
11.5. Checking and adjusting industrial television installations	222
11.6. Accepting industrial television installations for operation	229
Review questions	230
 Chapter 12. Installation and Set-Up of Electric Security and Fire-Alarm Systems	 231
12.1. General principles	231
12.2. Installing sensors and detectors	231
12.3. Installing receiving and monitoring devices and receiving stations	236

FOR OFFICIAL USE ONLY

12.4. Installing trunk lines for certain sensors and detectors . . .	237
12.5. Alignment and adjustment of alarm systems	242
12.6. Measuring electrical parameters of alarm systems	248
12.7. Final documentation and the turning-over of systems for operation	249
Review questions	251
Chapter 13. Installing and Adjusting Electric Clock Systems	251
13.1. General principles	251
13.2. Installing station facilities	251
13.3. Installing sympathetic clocks	254
13.4. Installing power supplies	256
13.5. Installing electric-clock lines	261
13.6. Checking electric-clock installations and putting them into operation	262
13.7. Accepting and turning over installation work	263
Review questions	265
Appendix 1. List of Basic Tools Used in the Installation of Industrial Communications Facilities	266
Appendix 2. Measuring Instruments Used to Adjust Industrial Communica- tion and Alarm Systems	267
Appendix 3. Types of Electric Industrial Communication and Alarm Systems	268
Appendix 4. Norms for the Duration of Automatic Telephone Exchange Installation	269
Appendix 5. List of Working Drawings for the Installation of Automatic Telephone Exchange K-100/2000	270
Appendix 6. Approximate List of Communication Equipment for Various Central Control Points at Construction and Installation Organizations .	270
Appendix 7. Norms for Engineering Plans of Communication Buildings . .	271
Appendix 8. Norms for the Consumption of Additional Materials	271
Appendix 9. List of Tools, Accessories, Implements and Materials Used in the Installation of Station Storage Batteries	272
Appendix 10. Some Data on S and SK-Type Station Acid Storage Batteries .	273
Appendix 11. Additional Materials for the Installation of Electric-Clock Systems	273

FOR OFFICIAL USE ONLY

3	Bibliography	274
	Subject Index	276

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9512

CSO: 1860/194

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NEW COLLECTION EXAMINES ELECTRONICS MATERIALS AND EQUIPMENT

Moscow TRUDY MOSKOVSKOGO ORDENA LENINA ENERGETICHESKOGO INSTITUTA, TEMATICHESKIY SBORNIK: MATERIALY I PRIBORY ELEKTRONNOY TEKHNIKI in Russian No 456, 1980 pp 2, 76-77

[Annotation and table of contents of book "Works of the Moscow Order of Lenin Power Engineering Institute, Collection Devoted to a Single Topic: Materials and Devices of Electrical Engineering", edited by Candidate of Technical Sciences and Docent V. A. Chizhov, Moskovskiy energeticheskiy institut, 77 pages]

[Text] The present collection includes articles by associates and graduate students in the Department of Electronic Devices. The articles reflect a number of trends in the scientific research conducted by the department: study of the vacuum and emission properties of electronic engineering materials; physical processes in electronic and semiconductor devices; the development of methods and equipment for ultrasonic flaw-detection and for optimization of information display systems. Workers in industry who are associated with the department through their scientific interests participate in a number of the articles.

The materials in the collection may be beneficial to workers at enterprises and NII [Scientific Research Institutes] of the electronics industry.

CONTENTS

	Page
Aksenov, V. P., Pitolin, A. I., Kachanov, V. K., Popko, V. P., Godin, A. B., Ryabov, G. Yu., Sokolov, I. V. Electrical engineering instruments and devices for generation, recording and conversion of ultrasonic oscillations	3
Anisimov, N. S., Lazarev, S. D., Merkulov, Ye. I., Peresleni, A. A. Electrical strength disturbance of vacuum microclearances	12
Vorob'yev, M. D., Smirnov, L. P., Grigor'yeva, T. P., Sliznev, S. M. Slow current fluctuations in a metal-dielectric-metal system	18
Rench, Ye. I. Optimization of heating conditions of planar thermocathodes	23

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Zhigarev, A. A., Kamunin, A. A. Method for designing systems formed by tapered electron beams	27
Dzhevala, A. P., Chizhov, V. A. A fluxmeter [potokomer] with variable conductivity	35
Chizhov, V. A., Mel'nikov, G. S., Dzhevala, A. P. Device for measuring the evacuation rate of small pumps	40
Smirnov, B. I. Apparatus for experimental research on the vacuum-tight properties of prestressed foil materials under complex heating and pulsed stressing conditions	43
Grigor'yev, V. N., Kravchenko, V. S., Obidin, G. I. Instrument for objective quality evaluation of visual information display systems	48
Gonzhalov, N. N., Peresleni, A. A., Titushina, V. P. Apparatus for studying the gas liberation kinetics of vacuum materials	51
Sherstnev, L. G., Yegorov, A. M., Vagina, T. A. Effectiveness of titanium for suppression of secondary emissions in electronic devices	56
Valygina, K. V., Obidin, G. I. On the question of determining optimal parameters of contrasting glasses for cathode-ray tube (CRT) screens	59
Stanotina, V. V. Electron microscopic study of lattices made from high-density pyrographites which are used in certain types of electronic devices	64
Chernyshev, V. A., Pantyukhov, S. T., Aleksenko, V. M. Dependence of an ionization current in gasses on pressure and intensity of an electric field	66
Sharikov, G. A., Degteva, V. Ye. Discharge of ZnS layers on an aluminum base by dark leakage and stimulated conductivity currents	71

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9194

CSO: 1860/214

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OPTIMIZATION AND CONTROL IN ELECTRONIC VACUUM PRODUCTION

Kiev OPTIMIZATSIYA I UPRAVLENIYE V ELEKTROVAKUUMNOM PROIZVODSTVE in Russian 1980
pp 2, 215

[Annotation and table of contents from book "Optimization and Control in Electronic Vacuum Production", by Ya. S. Podstrigach, Ya. I. Burak, V. I. Shelepets, S. F. Budz and A. B. Piontkovskiy, Izdatel'stvo "Naukova dumka", 216 pages]

[Text] The scientific bases and methods of optimization and control in electronic vacuum production are examined. These are based on systematic analysis, mathematical modeling of processes, formulation of appropriate extreme problems and formalization and algorithmization of control. The methodology which was developed is illustrated with examples from the most characteristic production processes and control subsystems.

The book is intended for scientific workers and engineers studying aspects of the optimization of systems with distributed parameters.

56 illustrations, 2 tables, bibliography--pp. 206-214 (183 titles).

CONTENTS

	Page
Preface	3
Chapter 1	
General principles of optimization and control	7
1. Operating principle and design features of receiver ELP [expansion unknown]	7
2. Structure of the typical production process	17
3. Production factors and quality indicators for the devices	27
4. Systematic approach to optimization of the technology	33
5. Aspects of the systematic control of the production process	39

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Chapter 2

Manufacture of the envelopes	44
1. Control of the glass composition and properties	44
2. Control of glass making	54
3. Control of preparation of the glass melt	67
4. Control of screen pressing	76
5. Studying the temperature field in the pressing process	79
6. An experimental-theoretical method for determining residual stresses	89
7. Optimization of screen annealing process	101
8. Optimization of the heat conditions for envelope fusing	116

Chapter 3

Application of coatings	121
1. Studying the process of luminophore precipitation	122
2. Control of the precipitation processes	134
3. Control of the application of mosaic screens	136
4. Destruction of organic films and heat treatment of aluminized screens	138

Chapter 4

Thermal vacuum treatment	147
1. Optimization of thermal outgasing conditions	148
2. Studying optimum thermal conditions	167
3. Optimization and study of conditions for outgasing color picture tubes	175

Appendix	181
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List of literature	206
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9194

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PHOTOELECTRONIC CONTROL SYSTEMS

Moscow FOTOELEKTRONNIYE SISTEMY UPRAVLENIYA in Russian 1980 (signed to press 26 Aug 80) pp 2, 204-5

[Annotation and table of contents from the book "Photoelectronic Control Systems", by Shimon Abramovich Vayner and Saveliy Abramovich Vayner, Izdatel'stvo "Mashinostroyeniye", 6000 copies, 208 pages]

[Text] This book is an examination of matters related to the theory and application of photoelectronic control systems(FESU). Included are investigations of the principles of control, mechanical trajectories, and the dynamic characteristics of such systems. Methods are presented for computation of the data which are represented by chart and graphic symbols. Descriptions are given of highly accurate self-adjusting FESU, high-speed adaptive FESU, systems which provide for equidistant correction during line tracing, and FESU employing delayed pulse feedback--all of which permit the optimal realization of detection and measurement of signal parameters and noise levels.

Examinations are made of contemporary foreign and domestic photoelectronic systems, various types of technical equipment employing FESU, the practical results of their use, and prospects for future applications.

The book is intended for use by technical engineering personnel involved in the development and application of equipment employing FESU as well as in the automation of technical processes.

Table of Contents

Foreword	3
Introduction	5
Chapter 1. Geometry of movement in FESU	8
1. Control of movement in FESU. Basic definitions	8
2. Laws of movement in FESU	13

FOR OFFICIAL USE ONLY

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3.	Controlling movement by the tractrix method	14
4.	Plotting straight lines by the tractrix method	19
5.	Plotting arcs and circles by the tractrix method	25
6.	Controlling movement by the three point method	28
7.	Plotting straight lines by the three point method	30
8.	Plotting circles by the three point method	32
9.	Plotting angles by the three point method	34
10.	Combined method of control using information relating to the displacement of the center of the field of view away from the plot line	37
11.	Plotting circles with the combined control method	39
12.	Plotting straight lines by the combined method	41
13.	Phasing and methods of automatically plotting equidistant trajectories in FESU	43
Chapter 2.	The Dynamics of FESU	46
1.	General nature of the dynamic characteristics of FESU	46
2.	Dynamic features of two-dimensional FESU drive systems	46
3.	Structural schemata of FESU using the tractrix method	53
4.	Stability of FESU using the tractrix method	58
5.	Dynamic errors in FESU	61
6.	Stability of FESU using the three point method	69
7.	Self-adjusting FESU	73
8.	FESU with integrated autotuning	75
9.	Testing FESU drive systems through the use of electronic models	77
Chapter 3.	Signals and Information Processing in FESU	85
1.	The chart as information carrier	85
2.	Computing graphic information in FESU	86
3.	Mathematical models of signals in FESU	88
4.	Determining the direction of the controlling vector	98
5.	Determining the midline of the chart according to maximal signal value	105
6.	Determining midline according to the position of the signal front	107
7.	Operative and long-term memory storage in FESU	109
8.	Positioning in FESU	111
9.	Special operating conditions of pulsed FESU	112
10.	Detecting signals and measuring their parameters in pulsed FESU	119
11.	Optimizing signal detection in FESU	121
12.	Optimizing measurement of signal parameters in pulsed FESU	122
Chapter 4.	The Elements of FESU	128
1.	Input devices used in FESU	128
2.	Signal processing methods in FESU and their implementation	135
3.	Rate selectors for variable speed electric motors	141

FOR OFFICIAL USE ONLY

4. Power amplifiers used in FESU	148
5. Control point setting and actuating mechanisms used in FESU	154
Chapter 5. Primary Functional Units of Contemporary FESU	160
1. A photoelectronic control system for the FOS-5[focusing and deflecting system] equipment	160
2. Changing the direction of line plotting in FESU	163
3. Establishing vector-integrated correction of errors	164
4. FESU with equidistant correction	168
5. Three-coordinate FESU	173
Chapter 6. Modern Technical Equipment Employing FESU	175
1. Comparative analysis of equipment using ChPU[numerical program control] and FESU	175
2. Milling machines employing FESU	180
3. Lathes using FESU	187
4. FESU in high-temperature sheet cutting machinery	191
5. Die surfacing machinery	197
6. Graphoanalytical converters	199
Conclusion	201
Bibliography	202

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9481
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SEMI-MARKOV PROCESSES WITH A DISCRETE SET OF STATES

Moscow POLUMARKOVSKIYE PROTSESSY S DISKRETNYM MNOZHESTVOM SOSTOYANIY in Russian 1980 (signed to press 19 Sep 80) pp 2, 270-1

[Annotation and table of contents from the book "Semi-Markov Processes with a Discrete Set of States", by Dmitriy Sergeyevich Sil'vestrov, Izdatel'stvo "Sovetskoye radio", 5000 copies, 272 pages]

[Text] This book is devoted to an exposition of the theory of semi-Markov processes having a discrete (finite or denumerable) set of states. Various methods of representing semi-Markov processes and associated Markov processes are considered. The important class of functionals known as moments of initial acquisition is examined in detail. Considerable attention is given to random processes associated with semi-Markov processes: to regenerative processes with semi-Markov changeovers, as well as accumulation processes with semi-Markov changeovers. Investigations of limit and ergodic theorems for these processes are carried out. The basic classes of queueing systems which can be described within the framework of the theory of semi-Markov processes are thoroughly analyzed.

The book is intended for use by engineers and mathematicians engaged in the design and operation of complex stochastic systems.

Table of Contents

Foreword	3
Chapter 1.	
Markov Chains	8
1.1. Introduction	8
1.2. Definition of Markov chains	9
1.3. Discrete homogeneous Markov chains	15
1.4. (J, x) -processes	22
1.5. Final remarks and conclusions	30

FOR OFFICIAL USE ONLY

FOR OFFICIAL USE ONLY

Chapter 2.	
Description of a Queueing System with the Use of Embedded (J, x)-Processes	33
2.1. Introduction	33
2.2. Queueing systems with unrestricted waiting line of the M/G/1 type	35
2.3. Closed renewable queueing systems of the M/G type	56
2.4. Semi-Markov queueing networks	74
2.5. Markov queueing systems	87
2.6. Final remarks and conclusions	92
Chapter 3.	
Definition of Semi-Markov Processes	94
3.1. Introduction	94
3.2. Definition of semi-Markov processes and the property of regularity	96
3.3. Associated Markov processes	99
3.4. Transition probabilities of associated Markov processes	109
3.5. Relationship of semi-Markov processes to time-homogenous Markov chains	120
3.6. Final remarks and conclusions	123
Chapter 4.	
Moments of Initial Acquisition of the Phase Space Domain by Means of Semi-Markov Processes	124
4.1. Introduction	124
4.2. Distribution of the moments of initial acquisition	127
4.3. Mathematical expectations for moments of initial acquisition	136
4.4. Sums of random variables and random vectors defined within a Markov chain up to the moment of initial acquisition	157
4.5. Final remarks and conclusions	163
Chapter 5.	
Processes with Semi-Markov Changeovers	165
5.1. Introduction	165
5.2. Regenerative processes	166
5.3. Processes with semi-Markov changeovers	182
5.4. Markov processes with semi-Markov changeovers	196
5.5. Structure of processes with semi-Markov changeovers, which describe queueing systems of the M/G type	203
5.6. Moments of initial acquisition for processes with semi-Markov changeovers	208
5.7. Final remarks and conclusions	217
Chapter 6.	
Limit and Ergodic Theorems for Processes with Semi-Markov Changeovers	218
6.1. Introduction	218

FOR OFFICIAL USE ONLY

6.2. Accumulation processes with semi-Markov changeovers	225
6.3. Limit theorems for accumulation processes with semi-Markov changeovers	232
6.4. Ergodic theorems averaged for accumulation processes with semi-Markov changeovers	246
6.5. Renewal theorem	252
6.6. Individual ergodic theorems for processes with semi-Markov changeovers	255
6.7. Final remarks and conclusions	262
Bibliography	265
Subject index	268

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9481

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